

# SNR/BER

Mean  $\mu_x = \frac{1}{N} \sum_{n=1}^N x[n]$

Power  $P_x = \frac{1}{N} \sum_{n=1}^N x[n]^2$

often factor mean out

$\hat{P}_x = \frac{1}{N} \sum_{n=1}^N (x[n] - \mu_x)^2 = \sigma^2 = \text{var}$

Energy  $E_x = \sum_{n=1}^N x[n]^2$

$\hat{E}_x = \sum_{n=1}^N (x[n] - \mu_x)^2$

SNR =  $\frac{\tilde{P}_{\text{signal}}}{\tilde{P}_{\text{noise}}}$  in db =  $10 \log \left( \frac{\tilde{P}_{\text{signal}}}{\tilde{P}_{\text{noise}}} \right)$

Stationary - not affected by shifts

Ergodic - " " " time

Find  $P_x$ , then  $P_{\text{noise}}$

No noise  $P_N = 0 \cdot \text{SNR} = \infty \text{ BER} = 0$

Noise  $\sim \sigma N + \mu$

$P(\text{Noise} < t) = P(\sigma N + \mu < t)$

$= \Phi \left( \frac{t - \mu}{\sigma} \right)$

Sometimes  $\mu_x = \frac{1}{2} \cdot 1 + \frac{1}{2} \cdot 0 = \frac{1}{2}$

$\sigma_x^2 = \frac{1}{2} \cdot 0^2 + \frac{1}{2} \cdot 1^2 - \left(\frac{1}{2}\right)^2 = \frac{1}{4}$

BER =  $P(\text{noise} > 0.5 | 0) \cdot P(0) + P(\text{noise} < 0.5 | 1) \cdot P(1)$

$= \Phi \left( -\frac{1}{2\sigma} \right) \cdot \frac{1}{2} + \frac{1}{2} \left( \Phi \left( -\frac{1}{2\sigma} \right) \right)$

$= \Phi \left( -\sqrt{\text{SNR}} \right)$

SNR (db) =  $10 \log \left( \frac{P_{\text{signal}}}{\sigma^2} \right)$

So for BER, Get  $\sigma^2$  from here, plug from SNR

Intersymbol Interference ISI

- Eye diagram

- all possible cases on top of each other

for BER  $P(1|1)P(\text{error}|1) + P(0|0)P(\text{error}|0) + \dots$

$\Phi \left( \frac{\text{mid-voltage it will be}}{\sigma} \right)$

Vth to minimize error rate

# 6.02 Quiz 2

Ways to avoid errors

Packet = {#, msg, chk}

Hash to detect error

Checksum - Just add up bits

- easy for one error to offset another

$P(>1 \text{ error}) = 1 - P(\text{no error}) = 1 - (1 - \text{BER})^k$

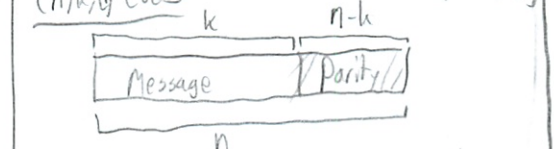
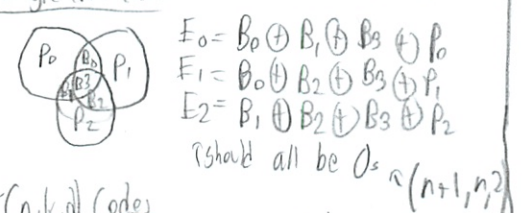
$P(2 \text{ errors}) = 1 - P(\text{no error}) - P(1 \text{ error})$

$= 1 - (1 - \text{BER})^k - k \cdot \text{BER} \cdot (1 - \text{BER})^{k-1}$

Hamming Distance = # digit positions wrong

$D = \text{min hamming dist} - \text{can correct } \left\lfloor \frac{D-1}{2} \right\rfloor$

Single error codes



$d = \text{Min Hamming distance can correct}$

$k/n = \text{code rate, } \leq 1, \text{ larger better}$

Rectangular } Need  $n \leq 2^{n-k} - 1$

Replicating (n, l, r) Good for code w/ Hamming = 2 = d

SECC (n, n-p, 3)  $n = 2^p - 1$  for  $p > 1$

Interleaving for burst errors

- Needs framing to know where to start

- 8b/10 works

Reed-Solomon

- the matrix, Galois field thing

- common (255, 223)

(n, k) Solves up to  $(n-k)/2$  errors

# Convolutional Codes

4/10

- always a fn of what came before

$P_i[n] = \left( \sum_{j=0}^{k-1} g_i[j] x[n-j] \right) \text{ mod } 2$

$g_i = k\text{-element generator polynomial for parity bit } P_i$

- either 1 or 0

- given

- interleave data parity for rate = 1/2

$= P_0[0]P_1[0]P_0[1]P_1[1] \dots$

message never transmitted

- like a SM or Trellise (over time)

- look for min hamming for most likely

then trace backward

Rec 11 10

00  $\rightarrow$  [2]  $\rightarrow$  [3]

01  $\rightarrow$  [1]  $\rightarrow$  [2]

10  $\rightarrow$  [2]  $\rightarrow$  [3]

11  $\rightarrow$  [1]  $\rightarrow$  [2]

path metric

branch metric

Hard decision - digitized 1 or 0

Soft decision

for each pt - like 0p

$\frac{(V_{p0}-0)^2 + (V_{p1}-0)^2}{\text{no } \sqrt{\dots}}$

k = # places using a bit - higher better

Protocols for Multi User

- high utilization  $U = \frac{\text{throughput all nodes}}{\text{max data rate of channel}}$

- fairness  $F = \frac{(\sum_{i=1}^N x_i)^2}{N \sum_{i=1}^N x_i^2}$   $\frac{1}{N} \leq F \leq 1$

- bounded wait

- scalability

TDMA - just give each person a slot

Aloha - p send packet

$U_{\text{slotted}} = N p (1-p)^{N-1}$

$P_{\text{best}} = \frac{1}{N}$  leads to  $U = \frac{1}{e} \approx 37\%$

can  $\downarrow$  p if collision

- if half =  $2^{-k} = \text{binary exp backoff}$

but bound  $p = \max(P_{\text{min}}, P/2)$

$P_{\text{min}} \ll 1/\max(N)$

Unslotted  $\geq \frac{N p (1-p)^{2N-1} N-1}{1/f} = TN p (1-p)^{2N-1}$

with large N  $V_{max} \approx \left(\frac{T}{2T-1}\right) \frac{1}{e}$

with large N.T  $U_{max} \approx \frac{1}{2e}$

$V_{max} = \frac{T}{2T-1} \left(1 - \frac{1}{(2T-1)N}\right)$

(Unslotted)  
 (carrier sense - even better)  
 - need random wait time after backoff  
 - inside a contention window

(CDMA - using orthogonal vectors)

Freq Division Multiplexing

- seq must be periodic (repeat)  
 $x[n] = x[n+N]$
- in N samples - fun freq =  $\frac{2\pi}{N}$
- harmonics possible  $k \cdot \frac{2\pi}{N}$
- negative means go other direction
- complex exponential

$e^{j\psi} = \cos(\psi) + j\sin(\psi)$

$\cos(\psi) = \frac{1}{2} e^{j\psi} + \frac{1}{2} e^{-j\psi}$

$\sin(\psi) = \frac{1}{2} e^{j\psi} - \frac{1}{2} e^{-j\psi}$

When  $\psi=0$   
 $e^{j0} = \cos(0) + j\sin(0) = 1 + 0j$

$\psi = \pm\pi$   
 $e^{j\pi} = e^{-j\pi} = -1$   
 $e^{j2\pi} = e^{-j2\pi} = (-1)^2 = 1$

Summing  
 $\sum_{n=\langle n \rangle} e^{jk \frac{2\pi}{N} n} = \begin{cases} N & k=0, \pm N, \pm 2N, \dots \\ 0 & \text{otherwise} \end{cases}$

Discrete Time Fourier Seq

$x[n] = \sum_{k=\langle k \rangle} a_k e^{jk \left(\frac{2\pi}{N}\right) n}$

k is over range of ints  
 • 0 for  $0 \leq \text{freq} < 2\pi$   
 •  $-N/2$  for  $-\pi \leq \text{freq} \leq \pi$

$a_k = \text{spectral coefficient (complex)}$

$a_k = \frac{1}{N} \sum_{n=\langle n \rangle} x[n] e^{-jk \left(\frac{2\pi}{N}\right) n}$

-  $k$ th spectral coefficient  
 $\uparrow$  = index = # cycles in N samples

Need to truncate/limit to a freq  
 - loses fine detail, like an avg

$H(e^{j\Omega}) = \sum_m h[m] e^{j\Omega m}$

$\delta[n] \rightarrow H(e^{j\Omega}) \rightarrow h[n]$

unit sample, periodic N times  
 $a_k = \frac{1}{N} x[0] e^{-jk \frac{2\pi}{N} \cdot 0} = \frac{1}{N}$

$h[n] = \sum_{k=\langle k \rangle} H(e^{jk \frac{2\pi}{N}}) e^{jk \frac{2\pi}{N} n}$

Finding the Os at  $\pm 4$

$H(e^{j\Omega}) = (e^{j\Omega})^2 - 2\cos(4)(e^{j\Omega}) + 1 = 0$

$h[0]=1 \quad h[1]=-2\cos(4) \quad h[2]=1$

Convolution in time domain  
 Multiplication in freq domain

poor man's low pass = convolve a bunch of Os together

$x_d \rightarrow \square \rightarrow y_d$  (Random)

$a_k e^{j\Omega k n} \rightarrow \square \rightarrow a_k H(e^{j\Omega k})$

$\Omega$  is the band you are assigned  
 All on 1 channel  
 have Os at start + end to bound

$\Omega = 2\pi f = 2\pi \frac{k}{N}$

$f_s$  = signal freq (cycles/sec)  
 $N$  = sample freq (samples/sec)

Filter to only certain freq  
 - as an LTI channel - choose  $h(c)$   
 - want 0 at many pts  
 - convolve many pts together

$h[n] = \delta[n] - \sqrt{2}\delta[n-1] + \delta[n-2]$

$H(e^{j\Omega}) = e^0 - \sqrt{2}e^{-j\Omega} + e^{-2j\Omega}$

$= 1 - \sqrt{2}\cos\Omega - \sqrt{2}\sin\Omega + \cos 2\Omega + \sin 2\Omega$

$= 1 - \sqrt{2}x + x^2$

if find 0, set = to 0, find x  
 $x = \cos\Omega + j\sin\Omega$ , find  $\Omega$

$C(\sqrt{2} \text{ here}) = \frac{1 + e^{-2j\Omega}}{-e^{-j\Omega}}$

$\frac{\cos}{a_p} = e^{j\psi} \cdot \frac{1}{2} + \frac{1}{2} e^{-j\psi}$  for  $0 \leq p \leq N-1$   
 or  $[-\frac{N}{2}] \leq p \leq [\frac{N}{2}]$

$\cos(p \left(\frac{2\pi}{N}\right) n)$  for  $-5 \leq p \leq 5$   
 works out to  $a_k = \begin{cases} \frac{1}{2} & k = \pm 1 \\ 0 & \text{otherwise} \end{cases}$

$a_p = \frac{1}{2} + e^{j\psi} = \frac{1}{2} + \cos\psi + j\sin\psi$

$a_{-p} = \frac{1}{2} + e^{-j\psi} = \frac{1}{2} + \cos\psi - j\sin\psi$

So  $x[n] = 1 + 2\cos(3 \frac{2\pi}{11} n) - 3\sin(5 \frac{2\pi}{11} n)$

$a_{13} = 2 \cdot \frac{1}{2} = 1$  (cos)

$a_{-5} = -3 \cdot \frac{1}{2} = -1.5j$

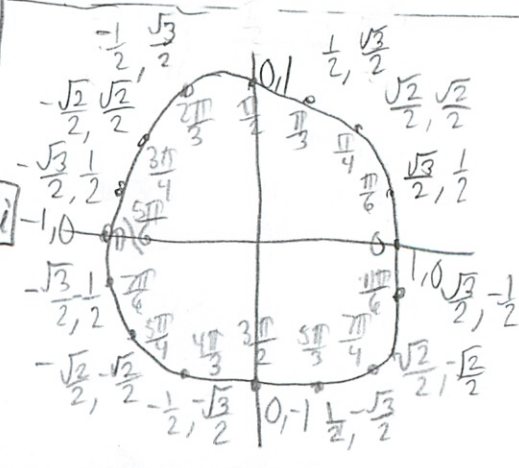
$a_5 = -3 \cdot \frac{1}{2} = 1.5j$

$a_k = 0$  otherwise

# states on state diagram =  $2^{k-1}$

$P(\text{noise} > 7.5) = P(\sigma N > 7.5) = P(W > \frac{15}{\sigma})$   
 $= P(N < \frac{-5}{\sigma}) = \Phi\left(\frac{-5}{\sigma}\right) = 1 - \Phi\left(\frac{5}{\sigma}\right)$

$(h,k)$  has at most  $2^{h-k}$  patterns  
 with min d =  $2^{k-1}$  can cover  $k$  errors  
 $k = \text{constraint length}$



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## 6.02 Tutorial Problems: Noise & Bit Errors

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### Problem 1.

Suppose the bit detection sample at the receiver is  $v + \text{noise}$  volts when the sample corresponds to a transmitted '1', and  $0.0 + \text{noise}$  volts when the sample corresponds to a transmitted '0', where  $\text{noise}$  is a zero-mean Normal(Gaussian) random variable with standard deviation  $\sigma_{\text{NOISE}}$ .

- A. If the transmitter is equally likely to send '0's or '1's, and  $v/2$  volts is used as the threshold for deciding whether the received bit is a '0' or a '1', give an expression for the bit-error rate (BER) in terms of the zero-mean unit standard deviation Normal cumulative distribution function,  $\Phi$ , and  $\sigma_{\text{NOISE}}$ .

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- B. Suppose the transmitter is equally likely to send zeros or ones and uses zero volt samples to represent a '0' and one volt samples to represent a '1'. If the receiver uses 0.5 volts as the threshold for deciding bit value, for what value of  $\sigma_{\text{NOISE}}$  is the probability of a bit error approximately equal to 1/5? Note that  $\Phi(0.85) \approx 4/5$ .

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- C. Will your answer for  $\sigma_{\text{NOISE}}$  in part (B) change if the threshold used by the receiver is shifted to 0.6 volts? Do not try to determine  $\sigma_{\text{NOISE}}$ , but justify your answer.

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- D. Will your answer for  $\sigma_{\text{NOISE}}$  in part (B) change if the transmitter is twice as likely to send ones as zeros, but the receiver still uses a threshold of 0.5 volts? Do not try to determine  $\sigma_{\text{NOISE}}$ , but justify your answer.

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### Problem 2.

Ben Bitdiddle is doing a 6.02 lab on understanding the effect of noise on data receptions, and is confused about the following questions. Please help him by answering them.

In these questions, assume that:

1. The sender sends 0 Volts for a "0" bit and 1 Volt for a "1" bit
2.  $P_{ij}$  = Probability that a bit transmitted as "i" was received as a "j" bit (for all four combinations of i and j, 00, 01, 10, 11)
3.  $\alpha$  = Probability that the sender sent bit 0
4.  $\beta$  = Probability that the sender sent bit 1
5. and, obviously,  $\alpha + \beta = 1$

The channel has non-zero random noise, but unless stated otherwise, assume that the noise has 0 mean and that it is a Gaussian with finite variance. The noise affects the received samples in an additive manner, as in the labs you've done.

- Which of these properties does the bit error rate of this channel depend on?
  - The voltage levels used by the transmitter to send "0" and "1"
  - The variance of the noise distribution
  - The voltage threshold used to determine if a sample is a "0" or a "1"
  - The number of samples per bit used by the sender and receiver

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- Suppose Ben picks a voltage threshold that minimizes the bit error rate. For each choice below, determine whether it's true or false.
  - $P_{01} + P_{10}$  is minimized for all alpha and beta
  - $\alpha * P_{01} + \beta * P_{10}$  is minimized
  - $P_{01} = P_{10}$  for all alpha and beta
  - if  $\alpha > \beta$  then  $P_{10} > P_{01}$
  - The voltage threshold that minimizes BER depends on the noise variance if  $\alpha = \beta$

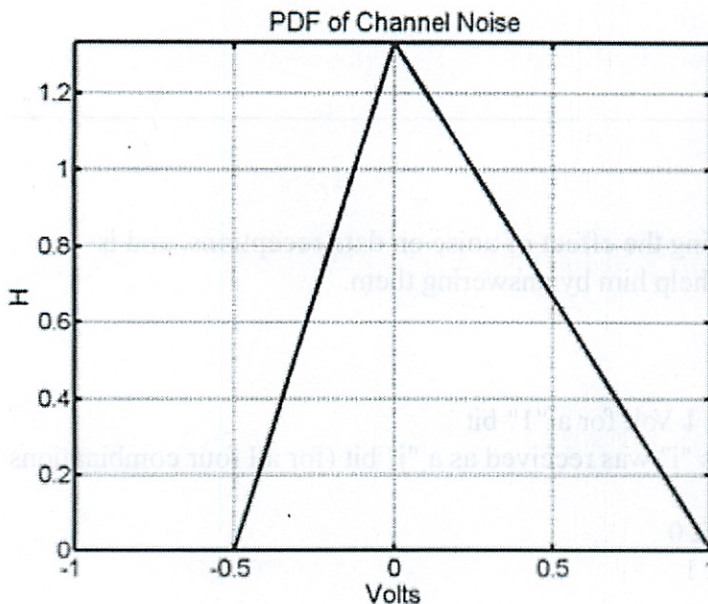
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- Suppose  $\alpha = \beta$ . If the noise variance doubles, what happens to the bit error rate?

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### Problem 3.

Messages are transmitted along a noisy channel using the following protocol: a "0" bit is transmitted as -0.5 Volt and a "1" bit as 0.5 Volt. The PDF of the total noise added by the channel,  $H$ , is shown below.

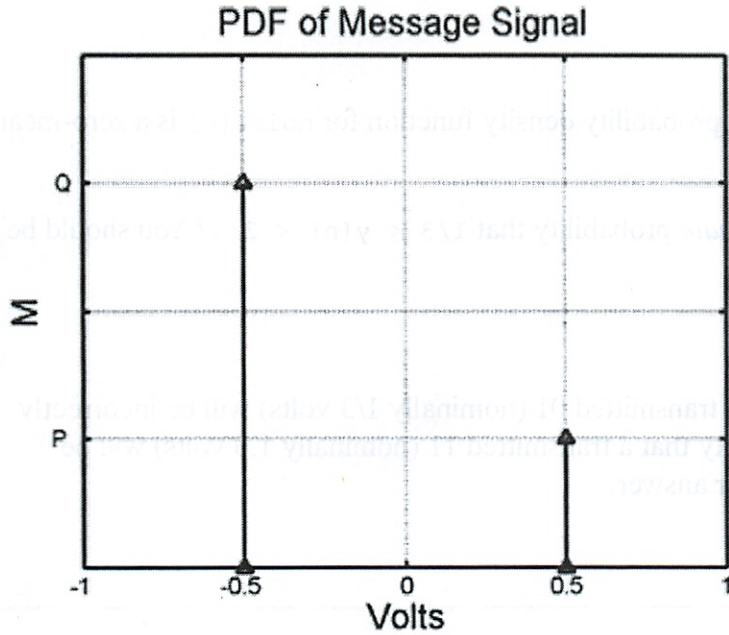


Not normal  
- pyramidal  
know

a. Compute  $H(0)$ , the maximum value of  $H$ .

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b. It is known that a "0" bits 3 times as likely to be transmitted as a "1" bit. The PDF of the message signal,  $M$ , is shown below. Fill in the values  $P$  and  $Q$ .



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c. If the digitization threshold voltage is 0V, what is the bit error rate?

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d. What digitization threshold voltage would minimize the bit error rate?

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**Problem 4.**

Consider a transmitter that encodes pairs of bits using four voltage values. Specifically:

- 00 is encoded as zero volts,
- 01 is encoded as  $(1/3)V_{\text{high}}$  volts,
- 10 is encoded as  $(2/3)V_{\text{high}}$  volts and
- 11 is encoded as  $V_{\text{high}}$  volts.

For this problem we will assume a wire that only adds noise. That is,

$$y[n] = x[n] + \text{noise}[n]$$

where  $y[n]$  is the received sample,  $x[n]$  the transmitted sample whose value is one of the above four voltages, and  $\text{noise}[n]$  is a random variable.

Please assume all bit patterns are *equally likely* to be transmitted.

Suppose the probability density function for  $\text{noise}[n]$  is a constant,  $\kappa$ , from  $-0.05$  volts to  $0.05$  volts and zero elsewhere.

A. What is the value of  $\kappa$ ?

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Suppose now  $V_{\text{high}} = 1.0$  volts and the probability density function for  $\text{noise}[n]$  is a zero-mean Normal with standard deviation  $\sigma$ .

C. If  $\sigma = 0.001$ , what is the *approximate* probability that  $1/3 < y[n] < 2/3$ ? You should be able to give a numerical answer.

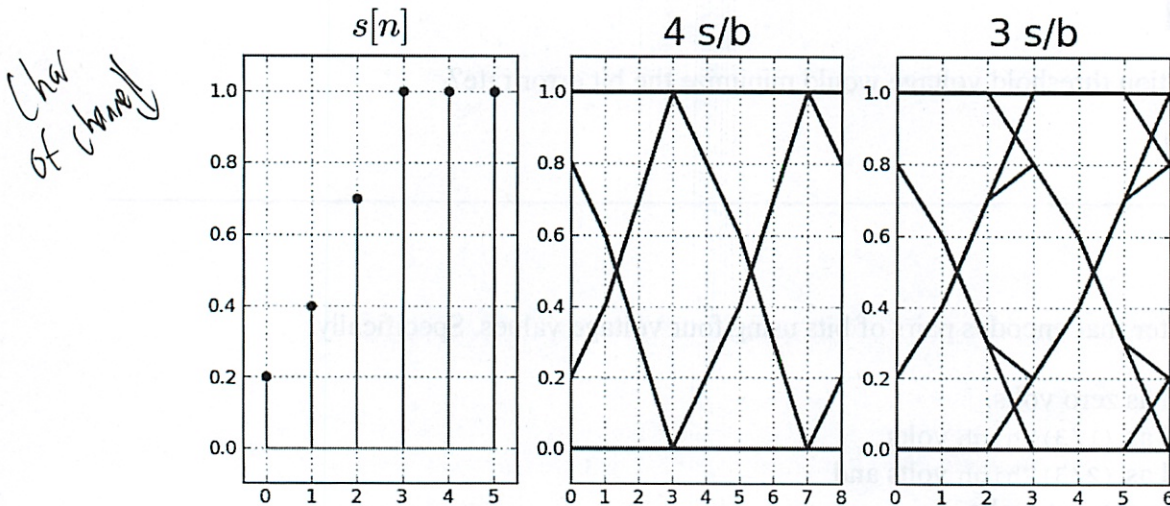
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D. If  $\sigma = 0.1$ , is the probability that a transmitted 01 (nominally  $1/3$  volts) will be incorrectly received the same as the probability that a transmitted 11 (nominally  $1.0$  volts) will be incorrectly received? Explain your answer.

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**Problem 5.**

Consider the figure below, which shows the step response for a particular transmission channel along with the eye diagram for channel response when transmitting 4 samples/bit and 3 samples/bit.



A. If the transmitter uses 3 samples/bit, under what conditions will it be possible to reliably (i.e., correctly) receive any sequence of transmitted bits?

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Suppose now that there is additive noise on this channel so that sometimes a transmitted bit is misidentified at the receiver. Let's investigate how the rate of bit errors is affected by changes in noise

probability density functions and number of samples per bit. In answering the questions below, please assume that the receiver uses the optimal detection sample for each bit (corresponding to the "center" of the eye) and uses a detection threshold of 0.5V.

- B. If we send 4 samples/bit down the noisy channel, the received voltage will be  $1.0 + \text{noise}$  when receiving a transmitted '1' bit, and  $0.0 + \text{noise}$  volts when receiving a transmitted '0' bit. If the noise is zero-mean Gaussian with standard deviation  $\sigma=0.25$ , what is the bit error rate? Assume that '0' and '1' bits are transmitted with a probability of 0.5, and that the noise is independent of the bit being transmitted.

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- C. If 3 samples/bit are used by the transmitter, the received voltage will be

$1.0 + \text{noise}$  or  $0.8 + \text{noise}$  when receiving a transmitted '1' bit, and

$0.2 + \text{noise}$  or  $0.0 + \text{noise}$  volts when receiving a transmitted '0' bit.

If the noise is **uniformly** distributed between the voltage values -1 and 1 volts, what is the bit error rate? Hint: are all four cases of received voltages equally likely?

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### Problem 6.

Suppose a channel has both noise and intersymbol interference, and further suppose the voltage at the receiver is:

$8.0 + \text{noise}$  volts when the transmitter sends a '1' bit preceded by a '1' bit

$6.0 + \text{noise}$  volts when the transmitter sends a '1' bit preceded by a '0' bit

$2.0 + \text{noise}$  volts when the transmitter sends a '0' bit preceded by a '1' bit

$0.0 + \text{noise}$  volts when the transmitter sends a '0' bit preceded by a '0' bit.

In answering the following parts, please assume the receiver uses 4.0 volts as the threshold for deciding the bit value.

- A. Suppose noise is Gaussian with standard deviation  $\sigma=1$  and all bit patterns are equally likely. Please determine the probability of a bit error.

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- B. Again suppose noise is Gaussian with standard deviation  $\sigma=1$ , and suppose that for a particular set of transmitted data, which we will refer to as checkerboard data, there is an increased probability of unequal contiguous bits. That is, for checkerboard data

$$p(00) = p(11) = 1/6$$

$$p(01) = p(10) = 1/3$$

What is the probability of bit error for the checkerboard case?

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## 6.02 Tutorial Problems: Error Correcting Codes

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**Problem 1.** For each of the following sets of codewords, please give the appropriate  $(n,k,d)$  designation where  $n$  is number of bits in each codeword,  $k$  is the number of message bits transmitted by each code word and  $d$  is the minimum Hamming distance between codewords. Also give the code rate.

A.  $\{111, 100, 001, 010\}$

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B.  $\{00000, 01111, 10100, 11011\}$

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C.  $\{00000\}$

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**Problem 2.** Suppose management has decided to use 20-bit data blocks in the company's new  $(n,20,3)$  error correcting code. What's the minimum value of  $n$  that will permit the code to be used for single bit error correction, i.e., that will achieve a minimum Hamming distance of 3 between code words?

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**Problem 3.** The Registrar has asked for an encoding of class year ("Freshman", "Sophomore", "Junior", "Senior") that will allow single error correction. Please give an appropriate 5-bit binary encoding for each of the four years.

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**Problem 4.** For any block code with minimum Hamming distance at least  $2t + 1$  between code words, show that:

$$2^{n-k} \geq 1 + \binom{n}{1} + \binom{n}{2} + \dots + \binom{n}{t}.$$

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**Problem 5.** Pairwise Communications has developed a block code with three data ( $D1, D2, D3$ ) and three parity bits ( $P1, P2, P3$ ):

$$P1 = D1 + D2$$

$$P2 = D2 + D3$$

$$P3 = D3 + D1$$

A. What is the  $(n,k,d)$  designation for this code.



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B. The receiver computes three syndrome bits from the (possibly corrupted) received data and parity bits:

$$E1 = D1 + D2 + P1$$

$$E2 = D2 + D3 + P2$$

$$E3 = D3 + D1 + P3.$$

The receiver performs maximum likelihood decoding using the syndrome bits. For the combinations of syndrome bits listed below, state what the maximum-likelihood decoder believes has occurred: no errors, a single error in a specific bit (state which one), or multiple errors.

$$E1 \ E2 \ E3 = 000$$

$$E1 \ E2 \ E3 = 010$$

$$E1 \ E2 \ E3 = 101$$

$$E1 \ E2 \ E3 = 111$$

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**Problem 6.** Dos Equis Encodings, Inc. specializes in codes that use 20-bit transmit blocks. They are trying to design a (20, 16) linear block code for single error correction. Explain whether they are likely to succeed or not.

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**Problem 7.** Consider the following (n,k,d) block code:

D0	D1	D2	D3	D4		P0
D5	D6	D7	D8	D9		P1
D10	D11	D12	D13	D14		P2
P3	P4	P5	P6	P7		

where D0-D14 are data bits, P0-P2 are row parity bits and P3-P7 are column parity bits. The transmitted code word will be:

D0 D1 D2 ... D13 D14 P0 P1 ... P6 P7

A. Please give the values for n, k, d for the code above.

**Show Answer**

B. If  $D_0 \ D_1 \ D_2 \ \dots \ D_{13} \ D_{14} = 0 \ 1 \ 0 \ 1 \ 0, \ 0 \ 1 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 0 \ 1$ , please compute P0 through P7.

**Show Answer**

C. Now we receive the four following code words:

$$M1: 0 \ 1 \ 0 \ 1 \ 0, \ 0 \ 1 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 0 \ 1, \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0$$

$$M2: 0 \ 1 \ 0 \ 1 \ 0, \ 0 \ 1 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 0 \ 1, \ 0 \ 0 \ 1 \ 1 \ 1 \ 0 \ 1 \ 0$$

$$M3: 0 \ 1 \ 0 \ 1 \ 0, \ 0 \ 1 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 0 \ 1, \ 1 \ 1 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0$$

$$M4: 0 \ 1 \ 0 \ 1 \ 0, \ 0 \ 1 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 0 \ 1, \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0$$

For each of received code words, indicate the number of errors. If there are errors, indicate if they are correctable, and if they are, what the correction should be.

**Show Answer**

**Problem 8.** The following matrix shows a rectangular single error correcting code consisting of 9 data bits, 3 row parity bits and 3 column parity bits. For each of the examples that follow, please indicate the correction the receiver must perform: give the position of the bit that needs correcting (e.g., D7, R1), or "no" if there are no errors, or "M" if there is a multi-bit uncorrectable error.

D1	D2	D3	R1
D4	D5	D6	R2
D7	D8	D9	R3
C1	C2	C3	—

None

1	1	1	1
1	1	0	0
0	1	1	0
0	1	1	—

0	0	1	0
0	1	1	1
0	1	1	0
0	0	1	—

1	1	0	0
1	0	0	1
0	0	1	0
0	0	1	—

1	0	1	0
1	0	0	1
1	1	0	0
1	1	1	—

1	1	1	0
1	1	1	1
1	0	0	0
0	0	1	—

Show Answer

**Problem 9.** Consider two convolutional coding schemes - I and II. The generator polynomials for the two schemes are

Scheme I:  $G_0 = 1101, G_1 = 1110$

Scheme II:  $G_0 = 110101, G_1 = 111011$

Notation is follows: if the generator polynomial is, say, 1101, then the corresponding parity bit for message bit  $n$  is

$$(x[n] + x[n-1] + x[n-3]) \text{ mod } 2$$

where  $x[n]$  is the message sequence.

- A. Indicate TRUE or FALSE
  - a. Code rate of Scheme I is 1/4.
  - b. Constraint length of Scheme II is 4.
  - c. Code rate of Scheme II is equal to code rate of Scheme I.
  - d. Constraint length of Scheme I is 4.

Show Answer

- B. How many states will there be in the state diagram for Scheme I? For Scheme II?

Show Answer

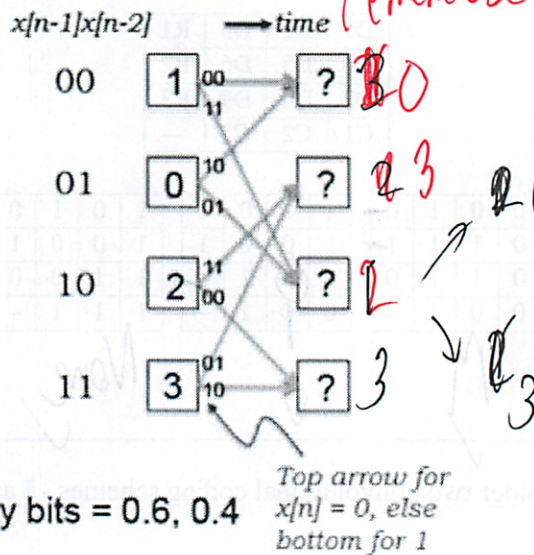
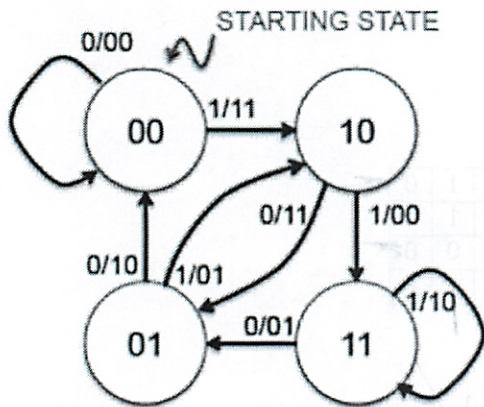
- C. Which code will lead to a lower bit error rate? Why?

Show Answer

- D. Alyssa P. Hacker suggests a modification to Scheme I which involves adding a third generator polynomial  $G_2 = 1001$ . What is the code rate  $r$  of Alyssa's coding scheme? What about constraint length  $k$ ? Alyssa claims that her scheme is stronger than Scheme I. Based on your computations for  $r$  and  $k$ , is her statement true?

Show Answer

**Problem 10.** Consider a convolution code that uses two generator polynomials:  $G_0 = 111$  and  $G_1 = 110$ . You are given a particular snapshot of the decoding trellis used to determine the most likely sequence of states visited by the transmitter while transmitting a particular message:



A. Complete the Viterbi step, i.e., fill in the question marks in the matrix, assuming a hard branch metric based on the Hamming distance between expected and received parity where the received voltages are digitized using a 0.5V threshold.

Show Answer

B. Complete the Viterbi step, i.e., fill in the question marks in the matrix, assuming a soft branch metric based on the square of the Euclidean distance between expected and received parity voltages. Note that your branch and path metrics will not necessarily be integers.

Show Answer

C. Does the soft metric give a different answer than the hard metric? Base your response in terms of the relative ordering of the states in the second column and the survivor paths.

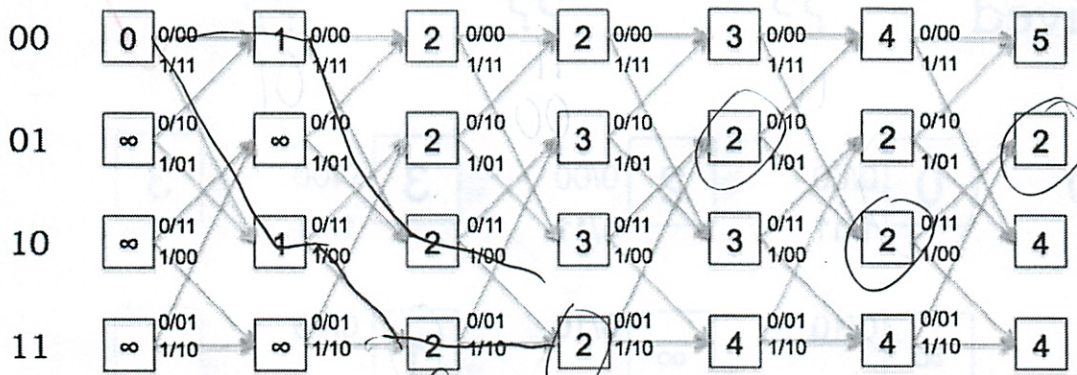
Show Answer

D. If the transmitted message starts with the bits "01011", what is the sequence of bits produced by the convolutional encoder?

Show Answer

The receiver determines the most-likely transmitted message by using the Viterbi algorithm to process the (possibly corrupted) received parity bits. The path metric trellis generated from a particular set of received parity bits is shown below. The boxes in the trellis contain the minimum path metric as computed by the Viterbi algorithm.

Time step	1	2	3	4	5	6
Received	01	01	00	01	01	11



E. Referring to the trellis above, what is the receiver's estimate of the most-likely transmitter state after processing the bits received at time step 6? *arbitrary*

Show Answer

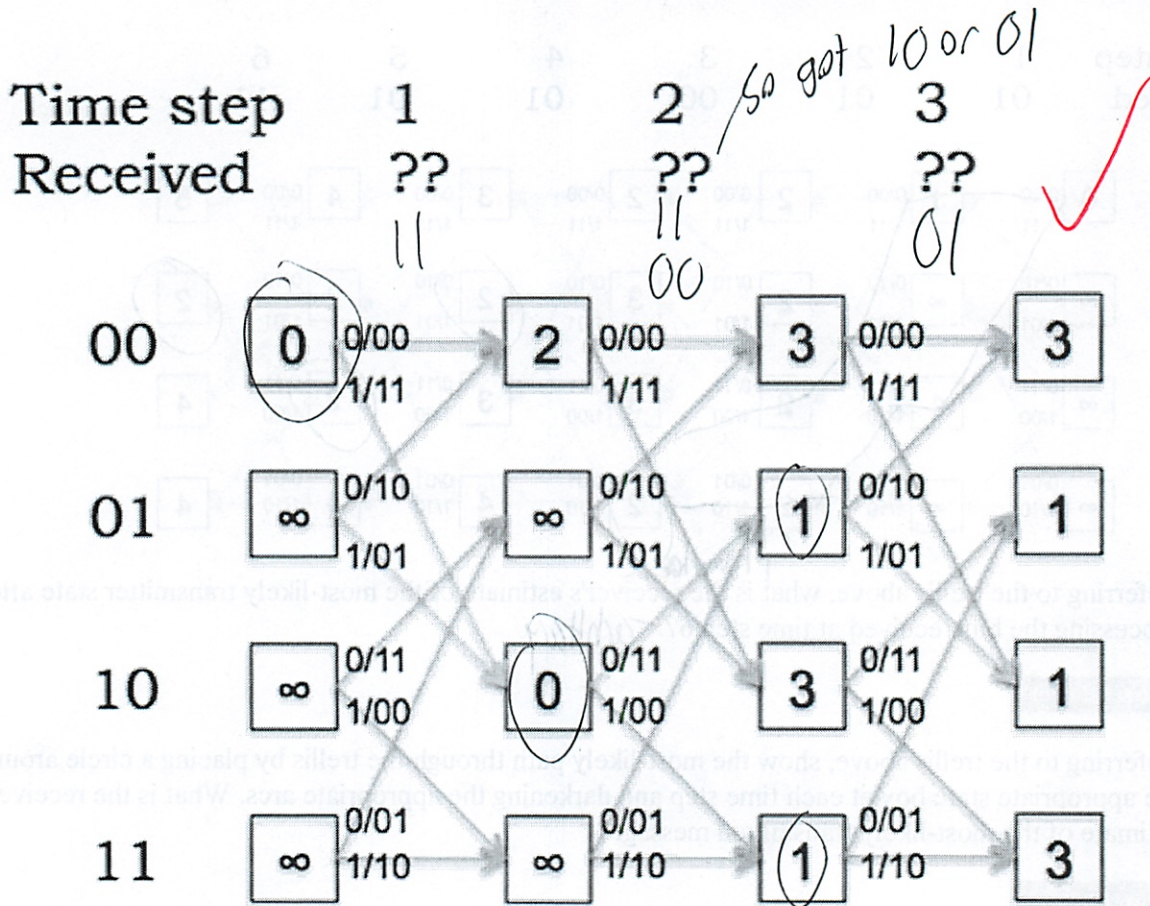
F. Referring to the trellis above, show the most-likely path through the trellis by placing a circle around the appropriate state box at each time step and darkening the appropriate arcs. What is the receiver's estimate of the most-likely transmitted message?

Show Answer

G. Referring to the trellis above, and given the receiver's estimate of the most-likely transmitted message, at what time step(s) were errors detected by the receiver? Briefly explain your reasoning.

Show Answer

H. Now consider the path metric trellis generated from a *different* set of received parity bits.



Referring to the trellis above, determine which pair(s) of parity bits could have been received at time steps 1, 2 and 3. Briefly explain your reasoning.

Show Answer

**Problem 11.** Consider a binary convolutional code specified by the generators (1011, 1101, 1111).

- A. What are the values of
  - a. constraint length of the code
  - b. rate of the code
  - c. number of states at each time step of the trellis
  - d. number of branches transitioning into each state
  - e. number of branches transitioning out of each state
  - f. number of expected parity bits on each branch

Show Answer

A 10000-bit message is encoded with the above code and transmitted over a noisy channel. During Viterbi decoding at the receiver, the state 010 had the lowest path metric (a value of 621) in the final time step, and the survivor path from that state was traced back to recover the original message.

- B. What is the likely number of bit errors that are corrected by the decoder? How many errors are likely left uncorrected in the decoded message?

Show Answer

- C. If you are told that the decoded message had no uncorrected errors, can you guess the approximate

number of bit errors that would have occurred had the 10000 bit message been transmitted without any coding on the same channel?

Show Answer

- D. From knowing the final state of the trellis (010, as given above), can you infer what the last bit of the original message was? What about the last-but-one bit? The last 4 bits?

Show Answer

Consider a transition branch between two states on the trellis that has 000 as the expected set of parity bits. Assume that 0V and 1V are used as the signaling voltages to transmit a 0 and 1 respectively, and 0.5V is used as the digitization threshold.

- E. Assuming hard decision decoding, which of the two set of received voltages will be considered more likely to correspond to the expected parity bits on the transition: (0V, 0.501V, 0.501V) or (0V, 0V, 0.9V)? What if one is using soft decision decoding?

Show Answer

**Problem 12.** Indicate whether each of the statements below is true or false, and a brief reason why you think so.

- A. If the number states in the trellis of a convolutional code is  $S$ , then the number of survivor paths at any point of time is  $S$ . Remember that if there is "tie" between to incoming branches (i.e., they both result in the same path metric), we arbitrarily choose only one as the predecessor.

Show Answer

- B. The path metric of a state  $s_1$  in the trellis indicates the number of residual uncorrected errors left along the trellis path from the start state to  $s_1$ .

Show Answer

- C. Among the survivor paths left at any point during the decoding, no two can be leaving the same state at any stage of the trellis.

Show Answer

- D. Among the survivor paths left at any point during the decoding, no two can be entering the same state at any stage of the trellis. Remember that if there is "tie" between to incoming branches (i.e., they both result in the same path metric), we arbitrarily choose only one as the predecessor.

Show Answer

- E. For a given state machine of a convolutional code, a particular input message bit stream always produces the same output parity bits.

Show Answer

**Problem 13.** Consider a convolution code with two generator polynomials:  $G_0=101$  and  $G_1=110$ .

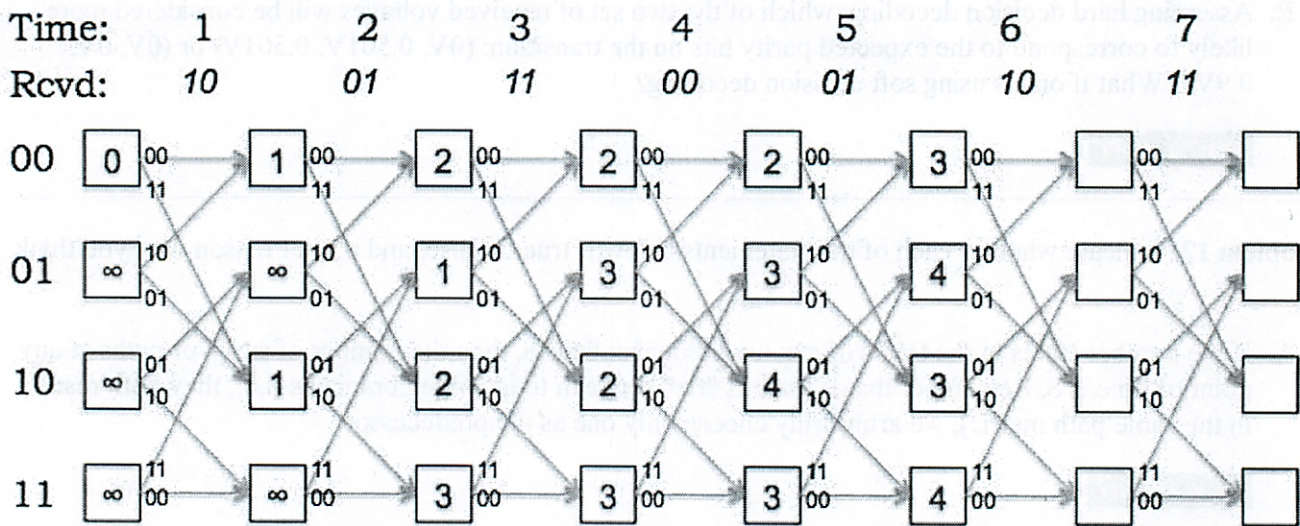
- A. What is code rate  $r$  and constraint length  $k$  for this code?

Show Answer

B. Draw the state transition diagram for a transmitter that uses this convolutional code. The states should be labeled with the binary string  $x_{n-1} \dots x_{n-k+1}$  and the arcs labeled with  $x_n/p_0p_1$  where  $x[n]$  is the next message bit and  $p_0$  and  $p_1$  are the two parity bits computed from  $G_0$  and  $G_1$  respectively.

Show Answer

The figure below is a snapshot of the decoding trellis showing a particular state of a maximum likelihood decoder implemented using the Viterbi algorithm. The labels in the boxes show the path metrics computed for each state after receiving the incoming parity bits at time  $t$ . The labels on the arcs show the expected parity bits for each transition; the actual received bits at each time are shown above the trellis.



C. Fill in the path metrics in the empty boxes in the diagram above (corresponding to the Viterbi calculations for times 6 and 7).

Show Answer

D. Based on the updated trellis, what is the most-likely final state of the transmitter? How many errors were detected along the most-likely path to the most-likely final state?

Show Answer

E. What's the most-likely path through the trellis (i.e., what's the most-likely sequence of states for the transmitter)? What's the decoded message?

Show Answer

F. Based on your choice of the most-likely path through the trellis, at what times did the errors occur?

Show Answer

[Show All Answers](#)[Hide All Answers](#)

## 6.02 Tutorial Problems: MAC protocols

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### Problem 1.

Which of these statements are true for correctly implemented versions of stabilized unslotted Aloha, stabilized slotted Aloha, and Time Division Multiple Access (TDMA)? Assume that the slotted and unslotted versions of Aloha use the same stabilization method and parameters.

- A. When the number of nodes is large, unslotted Aloha has a lower maximum throughput than slotted Aloha.

[Show Answer](#)

- B. When the number of nodes is large and nodes transmit data according to a Poisson process, there exists *some* offered load for which the throughput of unslotted Aloha is higher than the throughput of slotted Aloha.

[Show Answer](#)

- C. TDMA has no packet collisions.

[Show Answer](#)

- D. There exists some offered load pattern for which TDMA has lower throughput than slotted Aloha.

[Show Answer](#)


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### Problem 2.

Binary exponential backoff is a mechanism used in some MAC protocols. Which of the following statements is correct?

- A. It ensures that two nodes that experience a collision in a time slot will *never* collide with each other when they each retry that packet.
- B. It ensures that two or more nodes that experience a collision in a time slot will experience a lower probability of colliding with each other when they each retry that packet.
- C. It can be used with slotted Aloha but not with carrier sense multiple access.
- D. Over short time scales, it improves the fairness of the throughput achieved by different nodes compared to not using the mechanism.

[Show Answer](#)



### Problem 3.

In the Aloha stabilization protocols we studied, when a node experiences a collision, it decreases its transmission probability, but sets a lower bound,  $p_{\min}$ . When it transmits successfully, it increases its transmission probability, but sets an upper bound,  $p_{\max}$ .

- A. Why would we set a lower bound on  $p_{\min}$  that is not too close to 0?

**Show Answer**

- B. Why would we set  $p_{\max}$  to be significantly smaller than 1?

**Show Answer**

- C. Let  $N$  be the average number of backlogged nodes. What happens if we set  $p_{\min} \gg 1/N$ ?

**Show Answer**

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### Problem 4.

Consider a shared medium with  $N$  nodes running the slotted Aloha MAC protocol without any backoffs. A "wasted slot" is one in which no node sends data. The fraction of time during which no node uses the medium is the "waste" of the protocol.

- A. If the sending probability is  $p$ , what is the waste? What are the smallest and largest possible values of the waste?

**Show Answer**

- B. If the Aloha sending probability,  $p$ , for each node is picked so as to maximize the utilization, what is the corresponding waste?

**Show Answer**

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### Problem 5.

True or false?

Assume that the shared medium has  $N$  nodes and they are always backlogged.

- A. In a slotted Aloha MAC protocol using binary exponential backoff, the probability of transmission will always eventually converge to some value  $p$ , and all nodes will eventually transmit with probability  $p$ .

**Show Answer**

- B. Using carrier sense multiple access (CSMA), suppose that a node "hears" that the channel is busy at time slot  $t$ . To maximize utilization, the node should not transmit in slot  $t$  and instead transmit the packet in the next time slot with probability 1.

**Show Answer**

- C. There is some workload for which an unslotted Aloha with perfect CSMA will not achieve 100% utilization.

**Show Answer****Problem 6.**

Eight Cell Processor cores are connected together with a shared bus. To simplify bus arbitration, Ben Bittiddle, young IBM engineer, suggested time-domain multiplexing (TDM) as an arbitration mechanism. Using TDM each of the processors is allocated a equal-sized time slots on the common bus in a round-robin fashion. He's been asked to evaluate the proposed scheme on two types of applications: 1) core-to-core streaming, 2) random loads.

- 1) Core-to-core streaming setup: Assume each core has the same stream bandwidth requirement.
- 2) Random loads setup: core 1 load = 20%, core 2 load = 30%, core 3 load = 10%, core 4 = 5%, core 5 = 1%, core 6 = 3%, core 7 = 1%, core 8 load = 30%

Help Ben out by evaluating the effectiveness (bus utilization) of TDM under these two traffic scenarios.

**Show Answer****Problem 7.**

Randomized exponential backoff is a mechanism used to stabilize contention MAC protocols. Which of the following statements is correct?

- A. It ensures that two nodes that experience a collision in a time-slot will *never* collide with each other when they each retry that packet.

**Show Answer**

- B. It ensures that two or more nodes that experience a collision in a time-slot will experience a lower probability of colliding with each other when they each retry that packet.

**Show Answer**

- C. It can be used with slotted Aloha but not with CSMA.

**Show Answer****Problem 8.**

Three users X, Y and Z use a shared link to connect to the Internet. Only one of X, Y or Z can use the link at a given time. The link has a capacity of 1Mbps. There are two possible strategies for accessing

the shared link:

- TDMA: equal slots of 0.1 seconds.
- "Taking turns": adds a latency of 0.05 seconds before taking the turn. The user can then use the link for as long as it has data to send. A user requests the link only when it has data to send.

In each of the following two cases, which strategy would you pick and why?

A. X, Y and Z send a 40KBytes file every 1sec.

**Show Answer**

B. X sends 80KBytes files every 1sec, while Y and Z send 10KBytes files every 1sec.

**Show Answer**

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### Problem 9.

Alyssa P. Hacker is setting up an 8-node broadcast network in her apartment building in which all nodes can hear each other. Nodes send packets of the same size. If packet collisions occur, both packets are corrupted and lost; no other packet losses occur. All nodes generate equal load on average.

Alyssa observes a utilization of 0.5. Which of the following are consistent with the observed utilization?

A. True/False: Four nodes are backlogged on average, and the network is using Slotted Aloha with stabilization, and the fairness is close to 1.

**Show Answer**

B. True/False: Four nodes are backlogged on average, and the network is using TDMA, and the fairness is close to 1.

**Show Answer**

Now suppose Alyssa's 8-node network runs the Carrier Sense Multiple Access (CSMA) MAC protocol. The maximum data rate of the network is 10 Megabits/s. Including retries, each node sends traffic according to some unknown random process at an average rate of 1 Megabit/s per node. Alyssa measures the network's utilization and finds that it is 0.75. No packets get dropped in the network except due to collisions.

C. What fraction of packets sent by the nodes (including retries) experience a collision?

**Show Answer**

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### Problem 10.

*Note: this problem is useful to review how to set up and solve problems related to Aloha-like access*

protocols, but the calculations shown in the answer are more complex than we would ask on a quiz.

Consider a network with four nodes, where each node has a dedicated channel to each other node. The probability that any node transmits is  $p$ . Each node can only send OR receive one packet at a time. What is the utilization of the network? Assume each packet takes 1 time slot. Assume each node has 3 queues -- one for each other node -- and that each is backlogged.

**Show Answer**

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### Problem 11.

Suppose that there are three nodes seeking access to a shared medium using slotted Aloha, where each packet takes one slot to transmit. Assume that the nodes are always backlogged, and that each has probability  $p_i$  of sending a packet in each slot, where  $i = 1, 2$  and  $3$  indexes the node. Suppose that we assign more the sending probabilities so that

$$p_1 = 2(p_2) \text{ and } p_2 = p_3$$

- A. What is the utilization of the shared medium?

**Show Answer**

- B. What are the probabilities that maximize the utilization and the corresponding utilization?

**Show Answer**

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### Problem 12.

Suppose that two nodes are seeking access to a shared medium using slotted Aloha with binary exponential backoff subject to maximum and minimum limits of the probability  $p_{\max} = 0.8$  and  $p_{\min} = 0.1$ . Suppose that both nodes are backlogged, and at slot  $n$ , the probabilities the two nodes transmit packets are  $p_1 = 0.5$  and  $p_2 = 0.3$ .

- A. What are the possible values of  $p_1$  at slot  $n+1$ ? What are the probabilities associated with each possible value?

**Show Answer**

- B. What are the possible values of  $p_2$  at slot  $n+1$ ? What are the probabilities associated with each possible value?

**Show Answer**

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**Problem 13.** Bluetooth is a wireless technology found on many mobile devices, including laptops, mobile phones, GPS navigation devices, headsets, and so on. It uses a MAC protocol called Time Division Duplex (TDD). In TDD, the shared medium network has 1 master and  $N$  slaves. You may assume that the network has already been configured with one device as the master and the others as slaves. Each slave has a unique identifier (ID) that serves as its address, an integer between 1 and  $N$ . Assume that no devices ever turn off during the operation of the protocol. Unless otherwise

mentioned, assume that no packets are lost.

The MAC protocol works as follows. Time is slotted and each packet is one time slot long.

In every *odd* time slot (1, 3, 5, ...,  $2t-1$ , ...), the master sends a packet addressed to some slave for which it has packets backlogged, in round-robin order (i.e., cycling through the slaves in numeric order).

In every *even* time slot (2, 4, 6, ...,  $2t$ , ...), the slave that received a packet from the master in the immediately preceding time slot gets to send a packet to the master, if it has a packet to send. If it has no packet to send, then that time slot is left unused, and the slot is wasted.

- A. Alyssa P. Hacker finds a problem with the TDD protocol described above, and implements the following rule in addition:

From time to time, in an odd time slot, the master sends a "dummy" packet addressed to a slave even if it has no other data packets to send to the slave (and even if it has packets for other slaves).

Why does Alyssa's rule improve the TDD protocol?

**Show Answer**

Henceforth, the term "TDD" will refer to the protocol described above, augmented with Alyssa's rule. Moreover, whenever a "dummy" packet is sent, that time slot will be considered a wasted slot.

- B. Alyssa's goal is to emulate a round-robin TDMA scheme amongst the  $N$  slaves. Propose a way to achieve this goal by specifying the ID of the slave that the master should send a data or dummy packet to, in time slot  $2t-1$  (note that  $1 \leq t \leq \infty$ ).

**Show Answer**

Henceforth, assume that the TDD scheme implements round-robin TDMA amongst the slaves. Suppose the master always has data packets to send only to an arbitrary (but fixed) subset of the  $N$  slaves. In addition, a (possibly different) subset of the slaves always has packets to send to the master. Each subset is of size  $r$ , a fixed value. Answer the questions below (you may find it helpful to think about different subsets of slaves).

- C. What is the *maximum possible* utilization of such a configuration?

**Show Answer**

- D. What is the *minimum possible* utilization (for a given value of  $r$ ) of such a configuration?

Assume that  $r > N/2$ . Note that if the master does not have a data packet to send to a slave in a round, it sends a "dummy" packet to that slave instead. A dummy packet does not count toward the utilization of the medium.

**Show Answer**

## Problem 14.

Recall the MAC protocol with contention windows. Here, each node maintains a contention window,  $W$ , and sends a packet  $t$  idle time slots after the current slot, where  $t$  is an integer picked uniformly in  $[1, W]$ . Assume that each packet is 1 slot long.

Suppose there are two backlogged nodes in the network with contention windows  $W_1$  and  $W_2$ , respectively. Assume that both nodes pick randomly from  $[1, W_1]$  and  $[1, W_2]$  *at the current time* and that  $W_1 \geq W_2$ . What is the probability that the two nodes will collide the next time they each transmit?

**Show Answer**

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### Problem 15.

Eager B. Eaver gets a new computer with two radios. There are  $N$  other devices on the shared medium network to which he connects, but each of the other devices has only one radio. The MAC protocol is slotted Aloha with a packet size equal to 1 time slot. Each device uses a fixed transmission probability, and only one packet can be sent successfully in any time slot. All devices are backlogged.

Eager persuades you that because he has paid for two radios, his computer has a moral right to get twice the throughput of any other device in the network. You begrudgingly agree. Eager develops two protocols:

*Protocol A:* Each radio on Eager's computer runs its MAC protocol independently. That is, each radio sends a packet with fixed probability  $p$ . Each other device on the network sends a packet with probability  $p$  as well.

*Protocol B:* Eager's computer runs a single MAC protocol across its two radios, sending packets with probability  $2p$ , and alternating transmissions between the two radios. Each other device on the network sends a packet with probability  $p$ .

- A. With which protocol, A or B, will Eager achieve higher throughput?

**Show Answer**

- B. Which of the two protocols would you allow Eager to use on the network so that his expected throughput is double any other device's?

**Show Answer**

[Show All Answers](#)[Hide All Answers](#)

## 6.02 Tutorial Problems: Frequency Domain & Filters

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### Problem 1.

Give an expression for the magnitude of a complex exponential with frequency  $\varphi$ , i.e.,  $|e^{j\varphi}|$ . Hint: it's a numeric value independent of  $\varphi$ .

[Show Answer](#)

### Problem 2.

A. Prove the validity of the following formula, often referred to as the *finite sum formula*:

$$\sum_{n=0}^{N-1} \alpha^n = \begin{cases} N & \alpha = 1 \\ \frac{1-\alpha^N}{1-\alpha} & \text{for any complex number } \alpha \neq 1 \end{cases}$$

[Show Answer](#)

B. Use the finite sum formula to show:

$$\sum_{n=\langle N \rangle} e^{jk \frac{2\pi}{N} n} = \begin{cases} N & k = 0, \pm N, \pm 2N, \dots \\ 0 & \text{otherwise} \end{cases}$$

[Show Answer](#)

### Problem 3.

Consider the following signals, which are periodic with period  $N$ . For each signal compute the spectral coefficients  $a_k$ . In 6.02, we usually choose  $N$  consecutive  $k$ 's starting with  $-N/2$ .

A.  $x[n] = 1 + \sin((2\pi/N)*n) + 3*\cos((2\pi/N)*n) + \cos(2*(2\pi/N)*n + \pi/2)$ .

[Show Answer](#)

B.  $x[n] = 5*\cos(6\pi n + \pi) + 7*\cos(3\pi n)$

[Show Answer](#)

C.  $x[n]$  is a square wave with period  $N=4$  with the following values:  $x[0]=1$ ,  $x[1]=1$ ,  $x[2]=0$ ,  $x[3]=0$ .

[Show Answer](#)

D.  $x[n] = \cos(2\pi n/3) * \sin(2\pi n/9)$

**Show Answer**

E.  $x[n]$  has exactly one non-zero value per period, i.e.,  $x[m] \neq 0$  for some  $m$  and 0 otherwise. Compute the *magnitude* of  $a_k$ .

**Show Answer**

**Problem 4.**

If  $x[n]$  is real, *even* (i.e.,  $x[n] = x[-n]$ ) and periodic with period  $N$ , show that all the  $a_k$  are real.

**Show Answer**

**Problem 5.**

Suppose you're given the spectral coefficients  $a_k$  for a particular periodic sequence  $x[n]$ . Compute the spectral coefficients  $b_k$  for  $w[n] = x[n-\alpha]$ , i.e.,  $x$  time shifted by  $\alpha$  samples, in terms of the  $a_k$ .

**Show Answer**

**Problem 6.**

Consider an LTI system characterized by the unit-sample response  $h[n]$ .

A. Give an expression for the frequency response of the system  $H(e^{j\Omega})$  in terms of  $h[n]$ .

**Show Answer**

B. If  $h[0]=1, h[1]=0, h[2]=1,$  and  $h[n]=0$  for all other  $n$ , what is  $H(e^{j\Omega})$ ?

**Show Answer**

C. Let  $h[n]$  be defined as in part B and  $x[n] = \cos(\varphi n)$ . Is there a value of  $\varphi$  such that  $y[n]=0$  for all  $n$  and  $0 \leq \varphi \leq \pi$ ?

**Show Answer**

D. Let  $h[n]$  be defined as in part B. Find the maximum magnitude of  $y[n]$  if  $x[n] = \cos(\pi n/4)$ .

**Show Answer**

E. Let  $h[n]$  be defined as in part B. Find the maximum magnitude of  $y[n]$  if  $x[n] = \cos(-(\pi/2)n)$ .

**Show Answer**

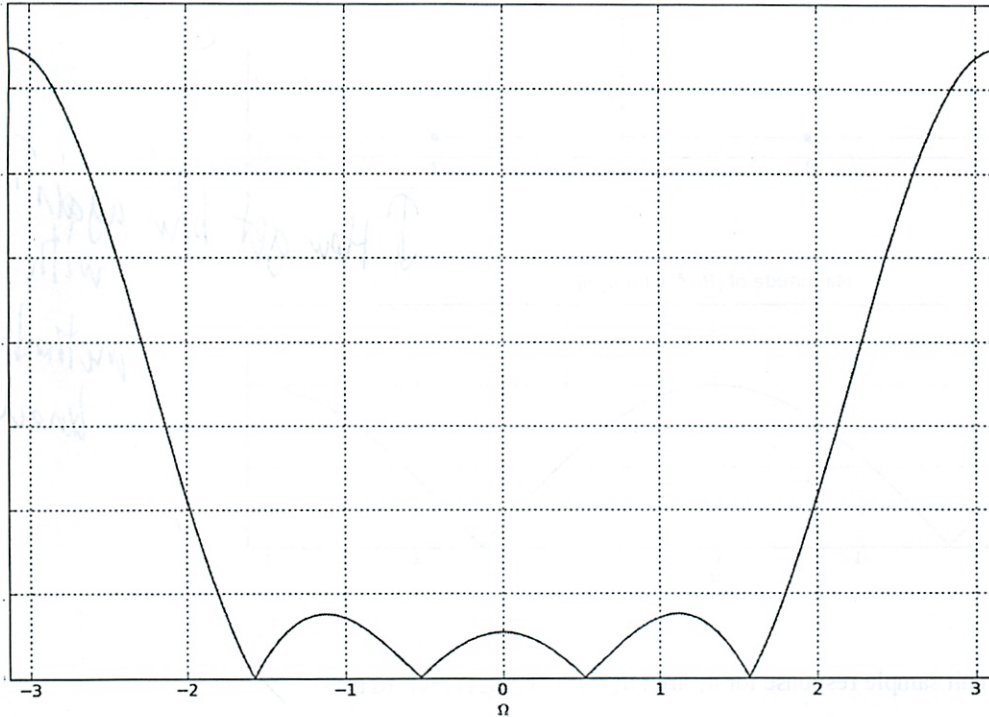
**Problem 7.**

In answering the questions below, please consider the unit sample response and frequency response of two filters,  $H_1$  and  $H_2$ ,

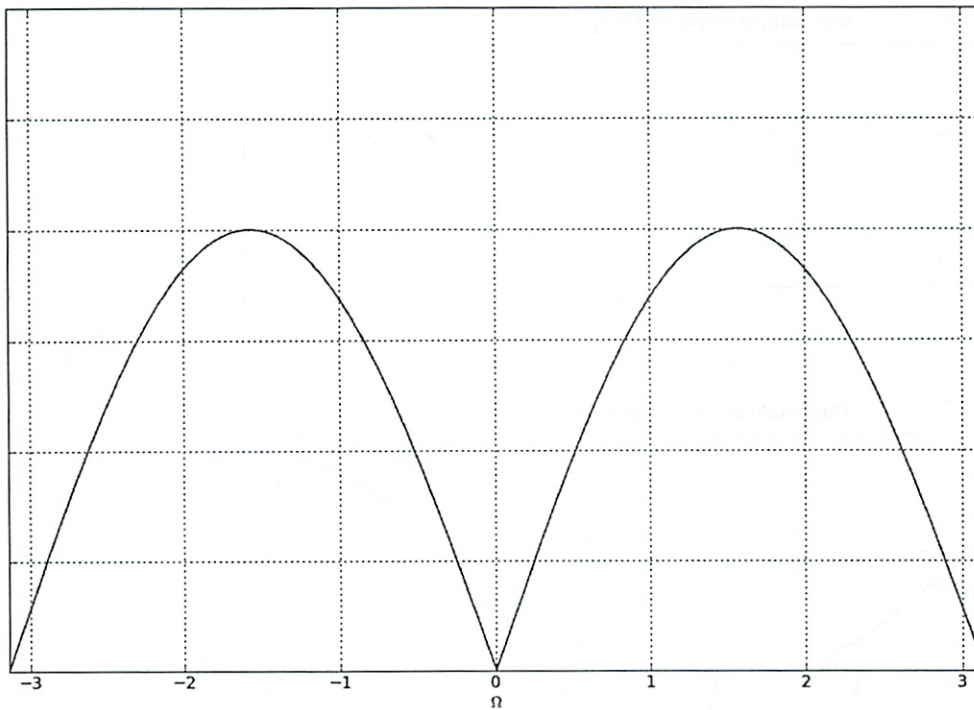


In answering the several parts of this review question consider four linear time-invariant systems, denoted A, B, C, and D, each characterized by the magnitude of its frequency response,  $|H_A(e^{j\Omega})|$ ,  $|H_B(e^{j\Omega})|$ ,  $|H_C(e^{j\Omega})|$ , and  $|H_D(e^{j\Omega})|$  respectively, as given in the plots below. This is a review problem, not an actual exam question, so similar concepts are tested multiple times to give you practice

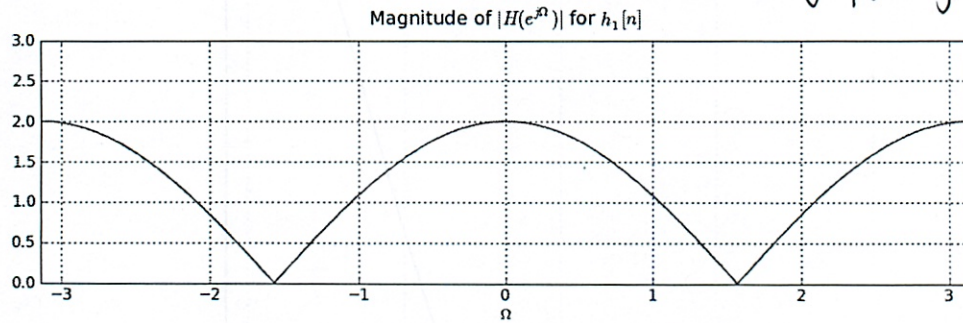
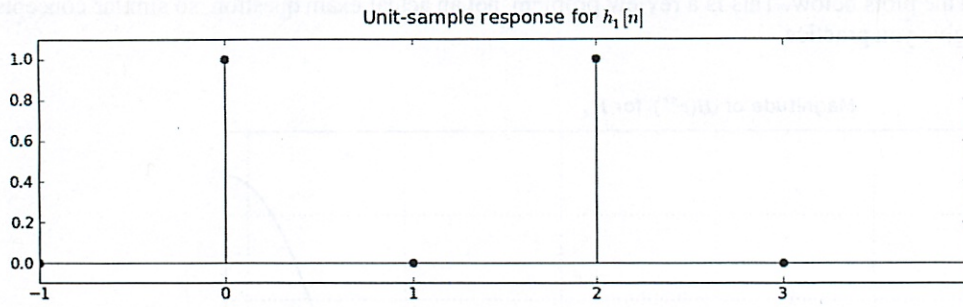
Magnitude of  $|H(e^{j\Omega})|$  for  $H_A$



Magnitude of  $|H(e^{j\Omega})|$  for  $H_B$

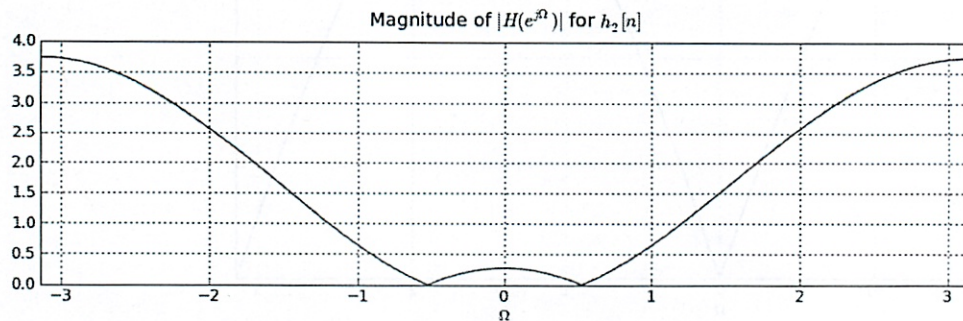
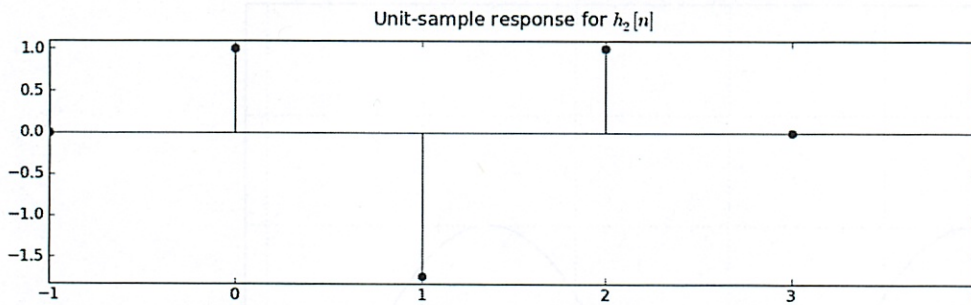


plotted below.



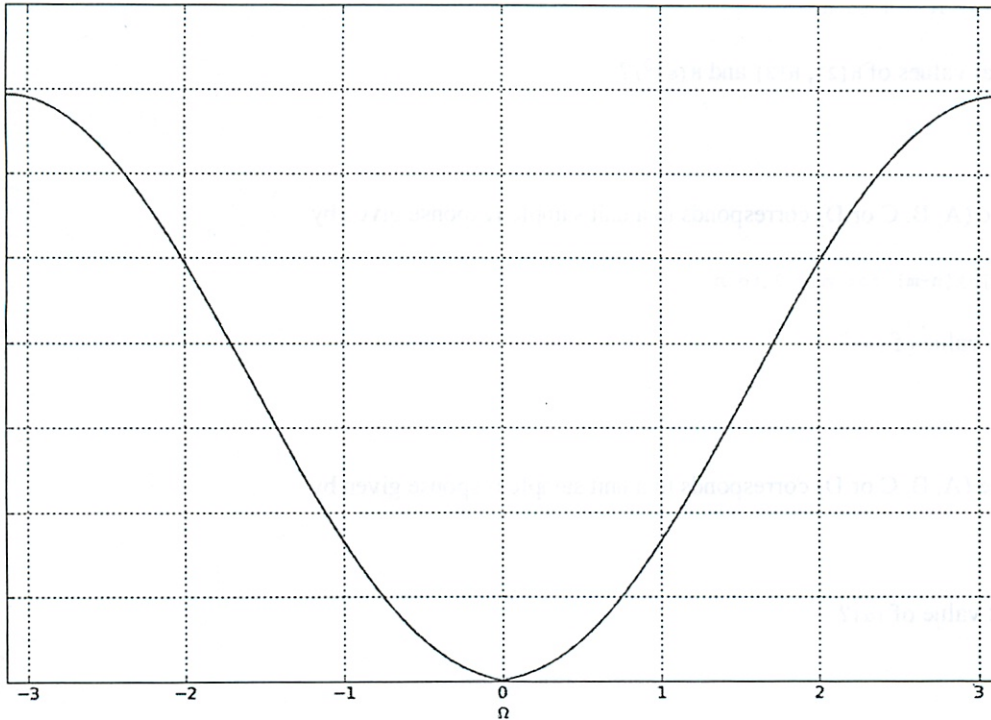
*How get b/w again with the method I know*

Note: the only nonzero values of unit sample response for  $h_1$  are :  $h_1[0] = 1, h_1[1]=0, h_1[2]=1$ .

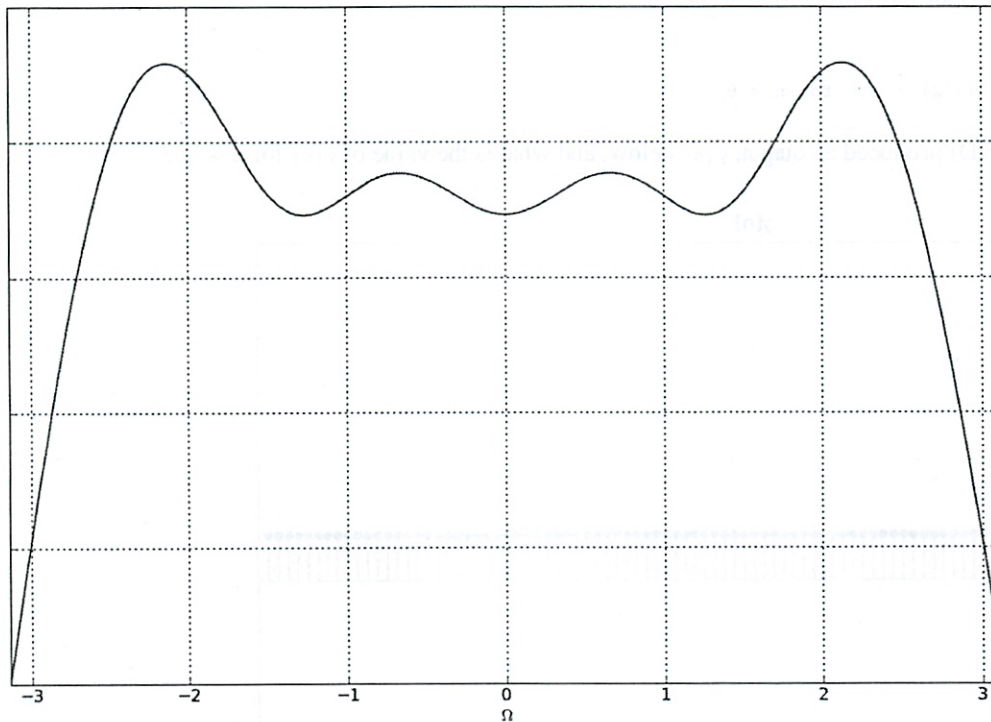


Note, the only nonzero values of unit sample response for  $h_2$  are :  $h_2[0] = 1, h_2[1]=-\sqrt{3}, h_2[2]=1$ .

Magnitude of  $|H(e^{j\Omega})|$  for  $H_c$



Magnitude of  $|H(e^{j\Omega})|$  for  $H_D$



A. Which frequency response (A, B, C or D) corresponds to a unit sample response given by

$$h[n] = \alpha \delta[n] - h_1[n]$$

and what is the numerical value of  $|\alpha|$ ?

**Show Answer**

B. Which frequency response (A, B, C or D) corresponds to a unit sample response given by

$$h[n] = \sum_m h_1[m]h_2[n-m] \text{ for } m = 0 \text{ to } n$$

and what are the numerical values of  $h[2]$ ,  $h[3]$  and  $H(e^{j0})$ ?

**Show Answer**

C. Which frequency response (A, B, C or D) corresponds to a unit sample response given by

$$h[n] = \alpha \delta[n] - \sum_m h_1[m]h_2[n-m] \text{ for } m = 0 \text{ to } n$$

and what is the numerical value of  $|\alpha|$ ?

**Show Answer**

D. Which frequency response (A, B, C or D) corresponds to a unit sample response given by

$$h[n] = \alpha \delta[n] - h_2[n]$$

and what is the numerical value of  $|\alpha|$ ?

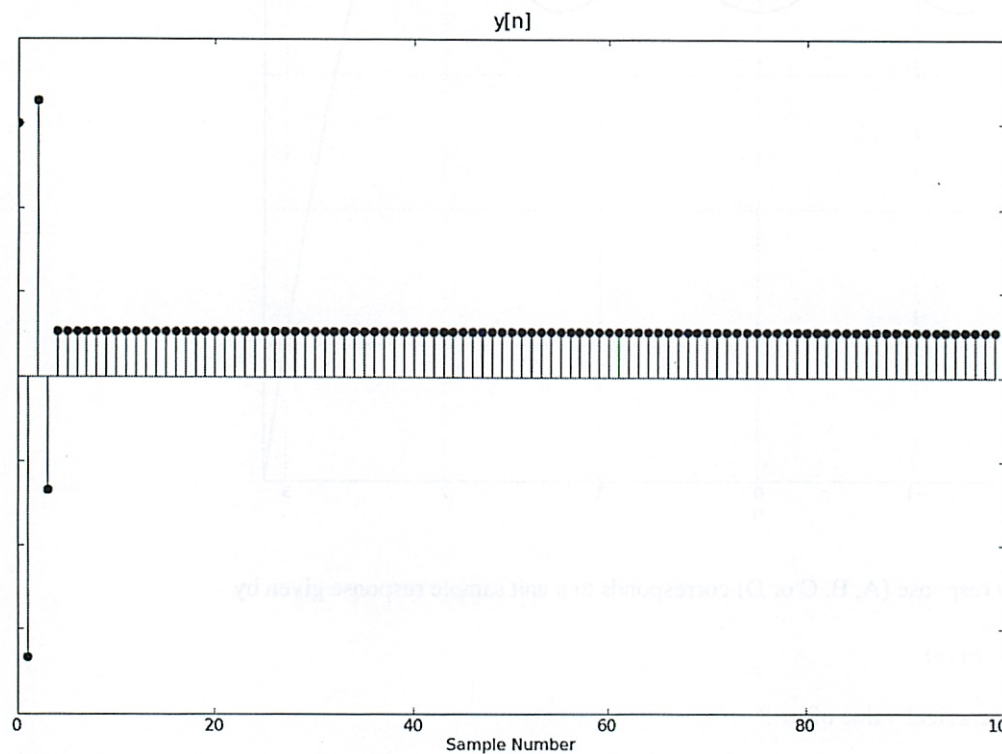
**Show Answer**

E. Suppose the input to each of the above four systems is

$$x[n]=0 \text{ for } n < 0 \text{ and}$$

$$x[n] = \cos(n\pi/6) + \cos(n\pi/2) + 1.0 \text{ for } n \geq 0$$

Which system (A, B, C or D) produced an output,  $y[n]$  below, and what is the value of  $y[n]$  for  $n > 10$ ?



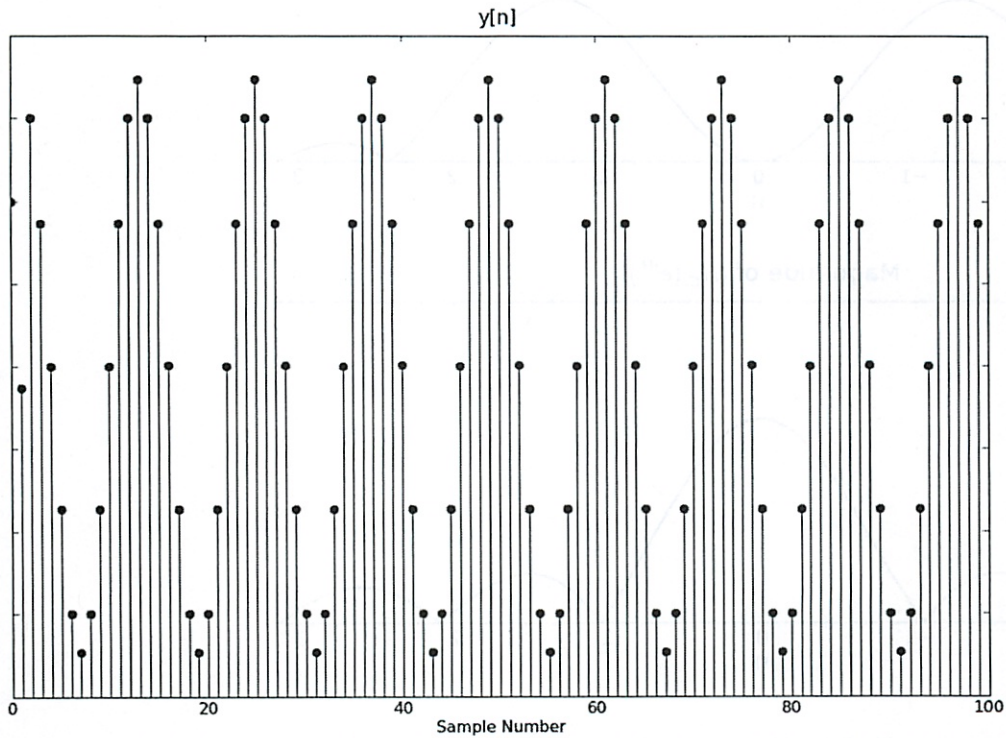
**Show Answer**

F. Suppose the input to each of the above four systems is

$$x[n]=0 \text{ for } n < 0 \text{ and}$$

$$x[n] = \cos(n\pi/6) + \cos(n\pi/2) + 1.0 \text{ for } n \geq 0$$

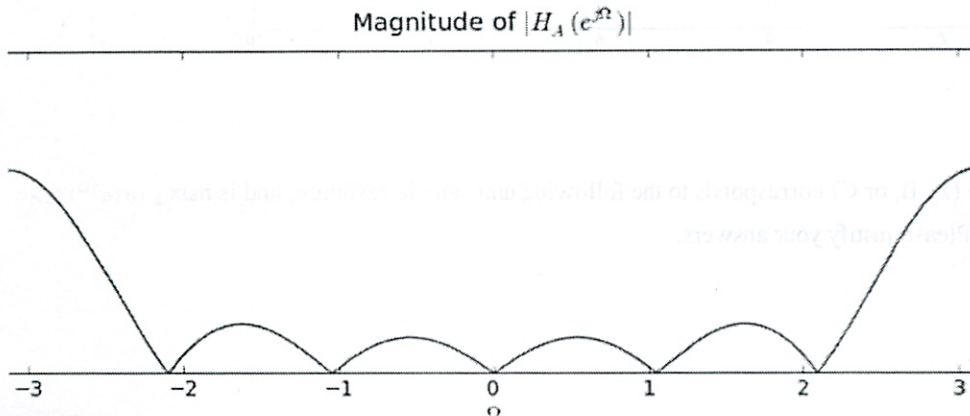
Which system ( $H_1$  or  $H_2$ ) produced an output,  $y[n]$  below, and what is the value of  $y[22]$ ?

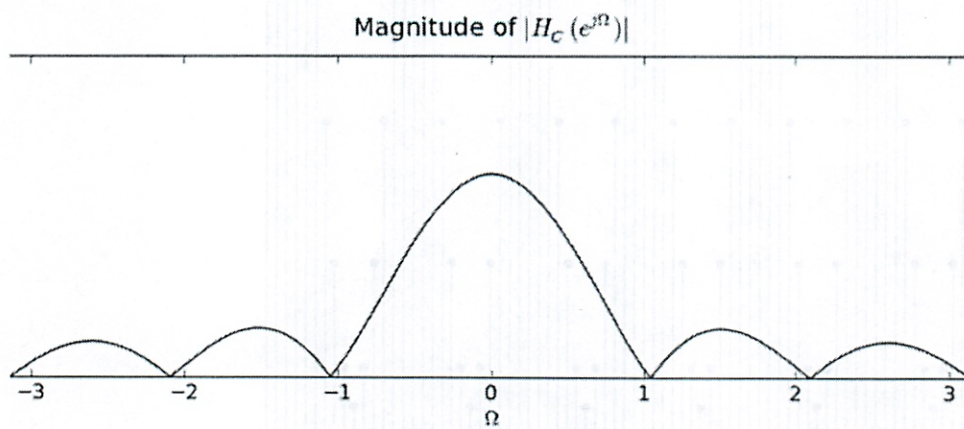
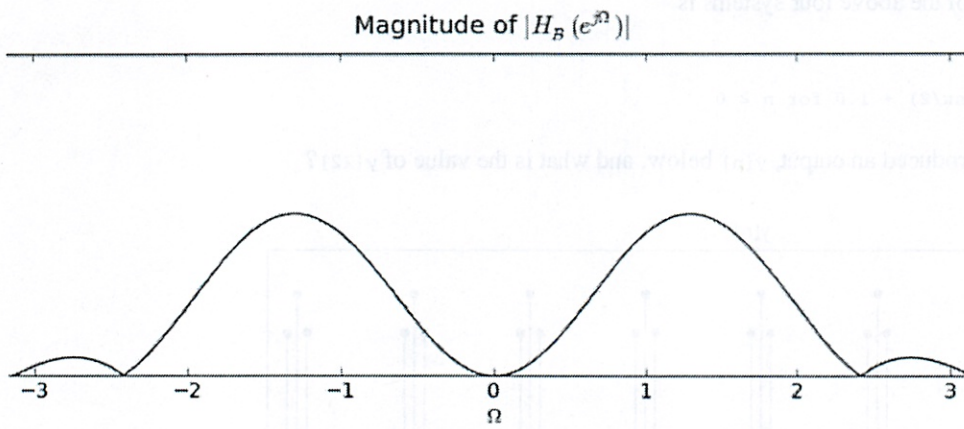


Show Answer

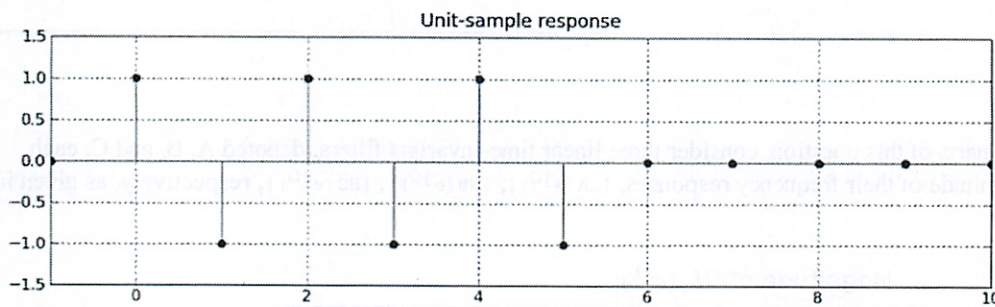
**Problem 8.**

In answering the several parts of this question, consider three linear time-invariant filters, denoted A, B, and C, each characterized by the magnitude of their frequency responses,  $|H_A(e^{j\Omega})|$ ,  $|H_B(e^{j\Omega})|$ ,  $|H_C(e^{j\Omega})|$ , respectively, as given in the plots below.



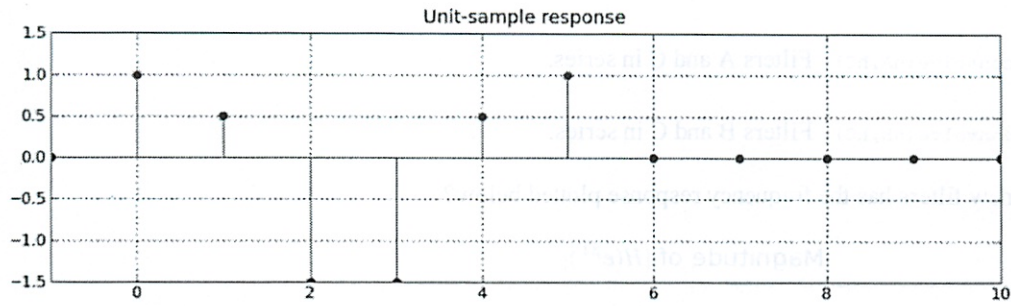


- A. Which frequency response (A, B, or C) corresponds to the following unit sample response, and what is  $\max_{\Omega} |H(e^{j\Omega})|$  for your selected filter? Please justify your selection.



**Show Answer**

- B. Which frequency response (A, B, or C) corresponds to the following unit sample response, and is  $\max_{\Omega} |H(e^{j\Omega})| > 6$  for your selected filter? Please justify your answers.

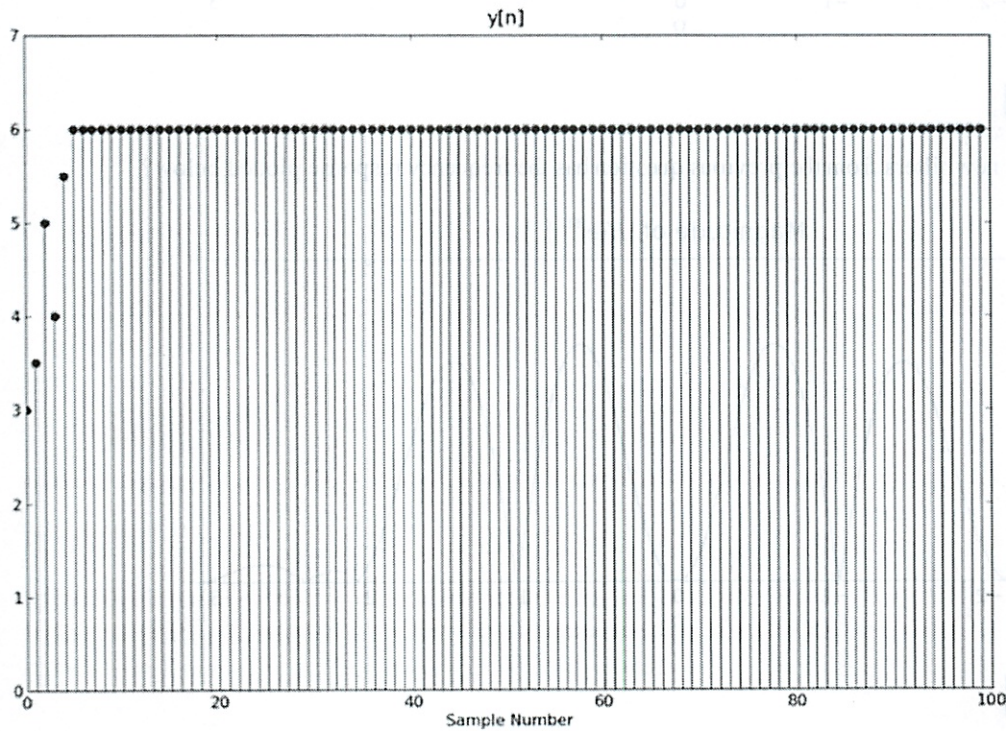


Show Answer

C. Suppose the input to each of the above three filters is  $x[n] = 0$  for  $n < 0$  and for  $n \geq 0$  is

$$x[n] = \cos\left(\frac{\pi}{3}n\right) + \cos(\pi n) + 1.0$$

Which filter (A, B, or C) produced the output,  $y[n]$  below, and what is  $\max_{\Omega} |H(e^{j\Omega})|$  for your selected system?



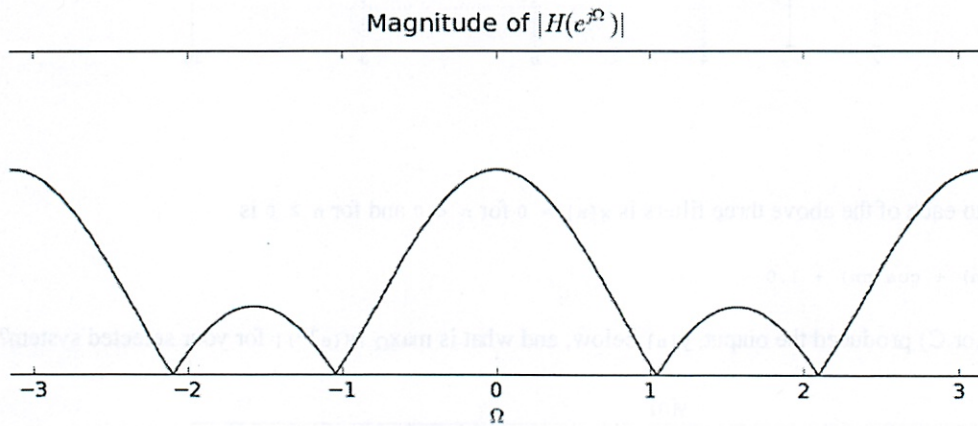
Show Answer

D. Six new filters were generated using the unit sample responses of filters A, B and C, denoted  $h_A[n]$ ,  $h_B[n]$ , and  $h_C[n]$  respectively. The unit sample responses of the new filters were generated in the following way:

- $h_1[n] = h_A[n] + h_B[n]$ . Filters A and B in parallel.
- $h_2[n] = h_A[n] + h_C[n]$ . Filters A and C in parallel.
- $h_3[n] = h_B[n] + h_C[n]$ . Filters B and C in parallel.
- $h_4[n] = \text{convolve}(h_A, h_B)$ . Filters A and B in series.

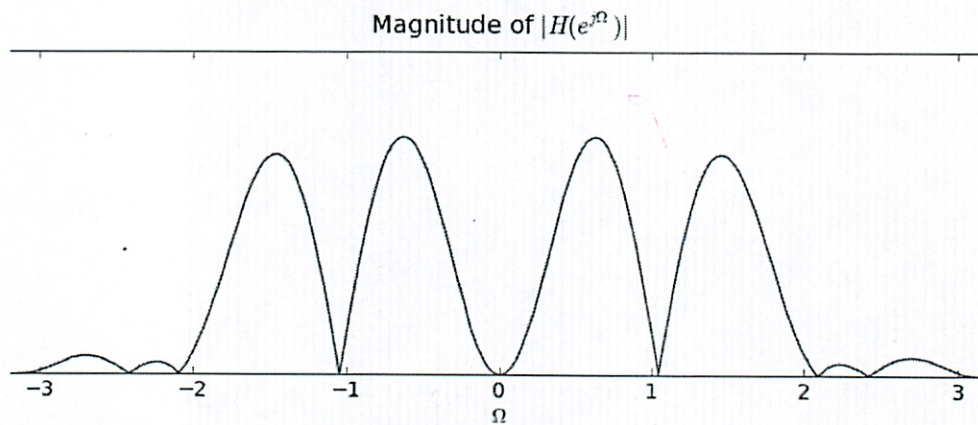
- $h_5[n] = \text{convolve}(h_A, h_C)$  . Filters A and C in series.
- $h_6[n] = \text{convolve}(h_B, h_C)$  . Filters B and C in series.

Which of the six new filters has the frequency response plotted below?



Show Answer

Which of the six new filters from the previous question has the frequency response plotted below?



Show Answer



# 6.02 Quiz 2 Review

4/11

## Noise + Bit errors

1.a.  $V + \text{noise}$   
 $0 + \text{noise}$

$$\text{noise} = N(0, \sigma_{\text{Noise}})$$

Give BER expression

$$\begin{aligned} \text{BER} &= P(\text{noise} > \frac{V}{2} | 0) P(0) + P(\text{noise} < \frac{V}{2} | 1) P(1) \\ &= \Phi\left(-\frac{1}{2\sigma}\right) \cdot \frac{1}{2} + \frac{1}{2} \left(\Phi\left(-\frac{1}{2\sigma}\right)\right) \\ &= \Phi\left(-\frac{1}{2\sigma}\right) \end{aligned}$$

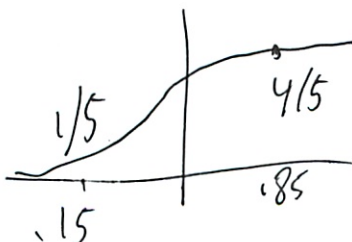
b.  $\frac{1}{5} = \Phi\left(-\frac{V_{th}}{2\sigma}\right)$  ← if not at ~~1~~ 1

How get by hand - solve for  $\sigma$

What is BER?  $\frac{\# \text{ errors}}{\# \text{ bits}}$  ?

But where does  $\Phi(0.85) = \frac{4}{5}$  come in

Unless its a match



$$0.15 = -\frac{1}{2\sigma}$$

2)

$-1.3 = \sigma$   
 $\tau$  but cant be -

$$BER = 1 - \Phi\left(\frac{1.5}{\sigma}\right)$$

if want BER = .2

I forgot the "1-" this step

$$\frac{1}{5} = 1 - \Phi\left(\frac{1.5}{\sigma_{noise}}\right)$$

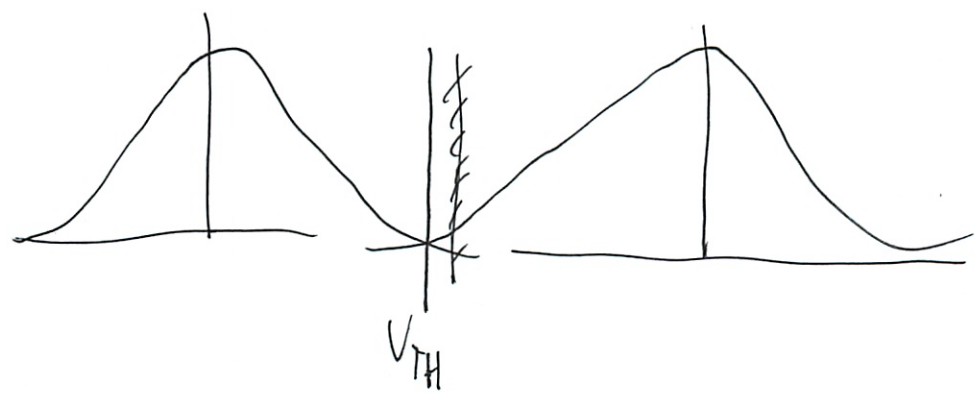
$$\Phi\left(\frac{1.5}{\sigma}\right) = 1 - \Phi\left(\frac{1.5}{\sigma}\right)$$

So  $\Phi\left(\frac{1.5}{\sigma}\right) = 0.9/5$

So  $.85 = \frac{1.5}{\sigma}$

So  $\sigma = .588$

c) ~~No~~ - Yes



Shifting  $\nearrow$  prob of bit error at  $\sigma$   
 So if prob bit error fixed,  $\sigma$  will  $\downarrow$  ✓

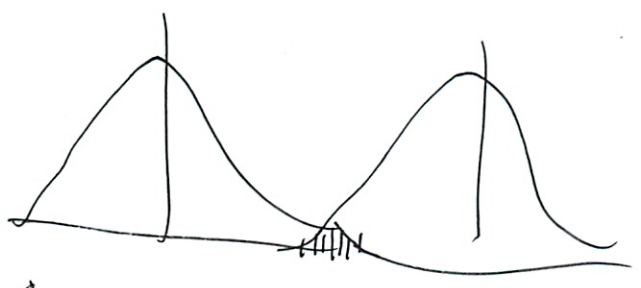
d) ~~Yes~~ Formula  $P(\text{error} | 0) P(0) + P(\text{error} | 1) P(1)$   
 $\uparrow$  same  $.33$   $\uparrow$  same  $.66$

actually still adds to same ✓ - no

3

~~11~~

7a) Isn't that fixed they said?  
I guess that varies while everything else is fixed



Noise + ISI

- a. yes move closer
- b. yes wider ✓
- c. yes where lie down
- d. no - this is just on samples level  
- later it will matter

- b.) a) true: only if  $\alpha = \beta$
- b)  $\alpha$  &  $\beta$  fixed, so ~~also~~ same as a ✓
- c) No
- d) No True review
- e) Yes No

c) Increases - but by how much?

4

3.a : 1.25 : ~~1~~

I guess they want you to do geometry

$$\frac{1}{2} \cdot .5 \cdot h + \frac{1}{2} \cdot 1 \cdot h = 1$$

$$\frac{3}{4}h = 1$$

$$\cdot \frac{4}{3}$$

$$h = \frac{4}{3} = 1.33$$



b) Same

0 3x likely

a = ~~0~~ P(0 trans)

b = ~~1~~ P(1 trans)

$$a = 3b$$

$$a + b = 1$$

$$a = 3(1-a)$$

$$a = 3 - 3a$$

$$4a = 3$$

$$a = \frac{3}{4}$$

$$b = \frac{1}{4}$$



↳ the fancy algebra way!  
- after this would not have been able to do

(5)

c) Signal + noise?

$$P(>0 | 0 \text{ trans})P(0) + P(<0 | 1 \text{ trans})P(1)$$

So to 0 look at PDF

$$\frac{1}{2} \circ 1.5 \text{ ---}$$

So need the blank

$$\frac{1.33}{1} = \frac{4/k}{1.5} = 1/3$$

$$\frac{1}{2} \circ \frac{1}{2} \circ \frac{1}{2} = \frac{1}{12}$$

$$\frac{1}{8}$$

look at 1 - assume noise just translated  
- to error prob here

~~1~~ ~~1~~

So all together

$$\frac{1}{2} \circ \frac{3}{4} = \frac{3}{8} = \text{BER}$$

They did something weird - adjusted it down

6

d) What would minimize error rate  
- think about - don't try to "solve"

Somewhere b/n 0, .5

Perhaps .25

That would be

$$\frac{1.33}{1} = \frac{4/3/4}{.25} \rightarrow \frac{4}{3} \cdot \frac{1}{4} = \frac{4}{12} = \frac{1}{3}$$

$$\frac{1.33}{.5} = \frac{4/3/2}{.25} \rightarrow \frac{4}{3} \cdot \frac{1}{2} = \frac{4}{6} = \frac{2}{3}$$

$$\frac{3}{4} \cdot \frac{1}{3} + \frac{1}{4} \cdot \frac{2}{3} = \frac{1}{4} + \frac{1}{6} = \frac{10}{24} = \frac{5}{12} = \text{BER}$$

Guess could generalize + "solve"

$$x = .3V$$

7) y. PDF noise is constant



a)  $k = \frac{1}{0.1}$



b) Now  $V_{\text{high}} = 1$  Noise <sup>Normal</sup>  $(0, \sigma)$

If  $\sigma = .001$  what is prop  $\frac{1}{3} \leq y < \frac{2}{3}$

~~Almost 0~~

~~No~~ 01 is  $\frac{1}{3}$

So its (half of the normal + half normal)  $\frac{1}{2}$

$$\frac{1}{2} \Phi\left(\frac{\frac{1}{2}}{.001}\right)$$

$\frac{1}{4}$  I don't think this is right

⑧

d) If  $\sigma = .1$

No -  $V_{TH} = .5$  - best place

So 01 more likely to be an error ✓

---

5. a) colors would be nice

$$V_{TH} = \frac{1}{2}$$

= prob

Yes eye still wide enough

if noise  $< .3 V$

just look at diagram

b) So now noise added

How is BER change?

- problem same as before or very similar

- ~~skipping~~ ~~secta~~ - how does it change w/ multiple?

= Oh C is where this is asked

- just add up  $\phi$  - check the likelihoods

01

10

think it's only 2

11

00



9

So that is =/ly likely

But is it Old ?  
etc

# 10) ECCs

1. Give  $(n, k, d)$ , rate

- But what exactly is transmitted here
- these are codewords
- but for what

$n = \#$  bits in codeword

$k = \#$  message bits trans. by each codeword

$d = \text{min hamming dist}$

150 ~~200~~

a)  $\{111, 100, 001, 010\}$

So  $n = 3$  since len 3 ✓

$k = 2$  since 4 of them ~~2~~  $2^2 = 4$

$d = 2$  hamming dist of them  
just look at

$$r = \frac{1}{3} \text{ bit} \left( \frac{k}{n} \right) = \frac{2}{3}$$

what is right.

b)  $\{00000, 01111, 10100, 11011\}$

$n = 5$

$k = 2$  since 4 codewords

$d = 2$

$$r = \frac{1}{5} \text{ bit} \left( \frac{k}{n} \right) = \frac{2}{5}$$

- don't really get what its asking

(11)

all ( ) and what in all world is this

trick

$n=5$

$k=0$

$d=und.$

rate = 0

no useful info transmitted

This was a stupid q

2. ~~What~~ 20 bit data blocks in  $(n, 20, 3)$   
What is min  $n$  so  $d=3$

$n \leq 2^{n-k} - 1$  is it? ✓

Min so

$n = 2^{n-20} - 1$

$n+1 = 2^{n-20}$

$n=22 \quad 23 = 2^3 \quad \otimes$

$n=23 \quad 24 = 2^4 \quad \times$

$n=25 \quad 26 = 2^5 \quad \otimes$   
 $32 \quad \otimes$  ← should work  
guess math error

but no  $d$  here!

$n$  does not matter

Oh the = does not apply  
- its just when it goes over the threshold

(2) Still a bit shaky on n, k, d thing

3. Registrar wants to encode class year w/ SEC  
So hamming distance of 2 or 3  
- well as much as possible

00000 ✓

~~11111~~

00111 ✓

11000

11011

11100

don't do this  
 $d = 2, 3$  can get 3

- but how do we know this  
must be 3!

4. ~~Can~~ For any block code w/  $d \leq 2t + 1$  show

$$2^{n-k} \geq 1 + \binom{n}{1} + \binom{n}{2} + \dots + \binom{n}{t}$$

How would you show this?

We never did this

$(n, k)$  has at most  $2^{n-k}$  patterns - must cover all error cases we wish to cover as well as no errors. With min  $d = 2t + 1$  can correct up to  $t$  errors. So can exp 0, 1, 2, ...  $t$  errors

as  $1 + \binom{n}{1} + \binom{n}{2} + \dots + \binom{n}{t}$

can't ~~exp~~  $> 2^{n-k}$

(3)

5. Pair wise

$$P_1 = D_1 + D_2$$

$$P_2 = D_2 + D_3$$

$$P_3 = D_3 + D_1$$

a)  $n =$  But what is sent?

$D_1 \ D_2 \ D_3 \ P_1 \ P_2 \ P_3$

$$k = 3$$

$$n = 6$$

$d =$  How to find again? list 8 possible code words + look  
 $\therefore 2$  b/c ~~1~~ wrong correctable  
 $\quad \quad \quad 2$  " not "  
 $\quad \quad \quad 3$  ← Actually look

b) Receiver

$$E_1 = D_1 + D_2 + P_1$$

$$E_2 = D_2 + D_3 + P_2$$

$$E_3 = D_3 + D_1 + P_3$$

What is max likelihood

a) No error ✓

(19)

b)  $E_2$  is wrong

$D_2 \quad D_3 \quad P_2$   
↑  
likely ✓

c)  $E_1 \quad E_2 \quad E_3 = 1 \quad 0 \quad 1$

↑  
 $D_1 \quad D_2 \quad P_1$       ↑  
 $D_3 \quad D_1 \quad P_3$   
↑  
likely ✓

d)  $E_1 \quad E_2 \quad E_3 = 1 \quad 1 \quad 1$

Multipl errors ✓

6. Want a  $(20, 16)$  linear block code

What is linear again?

- not on my cheatsheet lots of codes are linear?

$$20 = n \quad 16 = k$$

What does it mean to be single error correctable?

↑ must be able to uniquely identify  $n+1$  patterns

-  $n$  error patterns + 1 w/o errors.  
So  $2^{n-k} \geq n+1$ . Not true w/  $n=20 \quad k=16$  ← do the check 1st

(15)

8) So

$$1+1=0$$

$$1+1+1=1$$

$$1=1$$

$$0=0$$

On paper ✓

9) Conv scheme

Scheme 1 :  $G0 = 1101$      $G1 = 1110$

2     $G0 = 110101$      $G1 = 111011$

So notation  $1101$  means parity for  $n$  is

$$(x[n] + x[n-1] + 0 + x[n-3]) \pmod 2$$

↑ since bit

9) ~~True~~ False ✓

- 2 bits set for every message bit  $\frac{1}{2}$

- message bit never sent

I     $r = \frac{1}{2}$      $k = 4$

II     $r = \frac{1}{2}$      $k = 6$

↑ code length = constraint length

16) b) Constraint length  $n=2$  is 4

- What is constraint length?

I think just  $k$  - length of history generator looks at

False  $n$  is 6

c) True

d) True ✓

b) How many states?

- Oh I did not study state diagram

$2^4 - 1$

I) 8 II) 32

c) 2 - since more history ✓

d)  $r = \frac{1}{3}$  lower code rate  
 $k = 4$  - more redundancy  
So scheme stronger



(17)

10. The Vertebra

- will be last I do in this section

$$\begin{aligned}
 a. G0 &= 111 \\
 G1 &= 110
 \end{aligned}$$

Rec .6, .4  
 1, 0

start 00

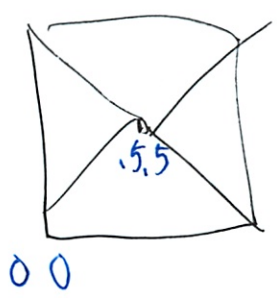
? So apply generator - well see  
 Go to 10

Why did I get wrong?

I know - I only did for 1st block

- this is not 1st time step!  
 - must look at other things

b) Grr - what is formula again



~~✓ thing~~  
 No ✓

$$\sqrt{(.5-0)^2 + (.5-0)^2}$$

(18) c) Did not fully do path firmed up

d)  $\begin{matrix} 0 & 1 & 0 & 1 & 1 \\ \underline{00} & 11 & 11 & 01 & 00 \end{matrix}$  ✓

e) Min path used state = 01  
Oh at 6

e) What is next step;

~~What~~ It does not receive anything

~~00~~ 10 ← 11 ← 01 ← 10 ← 01

f) Isn't there 2 paths? So 811010  
Yes

H) See paper ✓

(19)  
MAC

1.a.) Which are true?

Slotted  ~~$V = \frac{1}{e}$~~   $V = \frac{1}{2e}$

What is larger

$e \approx 3$  say

$\frac{1}{3}$  is larger

Slotted larger

True ✓

b) Is there some load - ~~year~~ when  
year when only 1 transmitter backed up / has  
something to transmit

~~True~~

False

~~True~~ slotted is always better  
than unslotted

c) True if done right ✓

- actually makes sense -  
less time to conflict

(20)

D) Oh yeah ✓

2. Binary exponential backoff. What is right

a) Not never, False

b) True - prob is lower ✓  
but lower than what?

c) False

d) Does nothing for fairness I think

---

3. a) So that one does not get so low it  
won't transmit again ✓

~~b) while~~

b) while don't want one to dominate - by  
always sending "capture effect" ✓

~~c) Min~~ c) Min to high - lots of collisions ✓

$U \sim 0$

26

4.a) Slotted aloha

IF  $p$ , what is waste?

$$U = Np(1-p)^{N-1}$$

$$(1-p)^N$$

So waste =  $1 - U$

oh something about them all together

$$\text{min} = 0 \quad \text{max} = 1$$

b)  $p_{\text{max}} = \frac{1}{N} \approx 37\%$

waste =  $1 - 37\% = 63\%$

$\frac{1}{e}$  but isn't  $U + \text{waste} = 1$ ?

5. True or false?

Nodes backlogged

or are they not counting collisions as waste

a) Will converge to  $p$  w/ all nodes w/ binary exponential backoff

No - will never converge ✓

b) Carrier sense - Transmit in next slot

No - wait random time ✓

(I feel like I get this section)

(22)

C) Some workload where not 100% util  
- are blank spots  
so true ✓

6. Process or cores

1. Streaming - very good 100% ✓

2. Random - some empty time

Do we need to calculate explicitly?

if any has load  $< \frac{100}{8}\%$  → waste

load  $> \frac{100}{8}\%$  → backed up

add  $\sum \max(12.5, \text{actual})$

could have easily have added ✓

23

- 7. Skip, similar
- 8. " "
- 9. " "
- 10. " complex
- 11. " similar
- 12. " "

13. a) I don't know why sending dummy packet helps

Unless there is something similar about it

Prevents slave from being starved - w/o slave w/ packets will never send packets

b) Yeah they said dummy = waste

Oh I get it now

So it sends dummy packet ~~up~~ when nothing to send.

did not lead close enough to see

(24)

# Freq Domain + Filters

This section I should know

$$1 = e^{j\ell} = \cos(\ell) + j \sin(\ell)$$

$$\text{add } \rightarrow \sqrt{\cos^2 \ell + \sin^2 \ell}$$

always = 1 I think ✓

2. a) prove

$$\sum_{n=0}^{N-1} \alpha^n = \begin{cases} N & \alpha = 1 \\ \frac{1-\alpha^N}{1-\alpha} & \text{for complex } \alpha \neq 1 \end{cases}$$

Case 1

So when  $\alpha = 1$

$$\begin{aligned} \sum_{n=0}^{N-1} 1^n &= 1^0 + 1^1 + 1^2 + \dots + 1^{N-1} \\ &= N \cdot 1 \\ &= N \end{aligned}$$

Case 2

$$\alpha^0 + \alpha^1 + \alpha^2 + \dots + \alpha^{N-1}$$

- same as 6.042 today

$$\alpha^0 = 1$$



25

I don't get the rest of it  
- Not in 6.042 either

For ~~all~~  $d \neq 1$

1. Expand the sum - What would this be  $1 + d + d^2 + \dots + d^{N+1}$

2. Multiply by  $\frac{1-d}{1-d}$  so its always this

3. Simplify the numerator

b) Use finite formula to show

$$\sum_{n=-N}^N e^{ik \frac{2\pi}{N} n} = \begin{cases} N & k=0, \pm N, \pm 2N \dots \\ 0 & \text{otherwise} \end{cases}$$

So it is 0 elsewhere

Adds up to balance at



Simply, use finite sum formula and  $e^{j\pi} = -1$

Oh tricky

So when it is 1?

- when it circles  $N$ ?

otherwise cancels out?

(20) (26 pages in - so sick of this)

3a) Compute  $a_k$

$$x[n] = 1 + \sin \frac{2\pi}{N} n + 3 \cos \left( \frac{2\pi}{N} n \right) + \cos \left( 2 \frac{2\pi}{N} n + \frac{\pi}{2} \right)$$

$$a_{+1} = -\frac{j}{2}$$

$$a_{-1} = \frac{j}{2}$$

$$a_1 = 3 \cdot \frac{1}{2}$$

$$a_{-1} = 3 \cdot \frac{1}{2}$$

$$a_{+2} = 2 \cdot \frac{1}{2}$$

what do you use  
gather terms  
add

?  
What does this do?  
I'm guessing offsets it  
So is like if sin did it

$a_k$  are periodic w/ period  $N$  ~~so  $2\pi/N$~~   
-oh just means it repeats

$$a_{-2} = -\frac{1}{2}j$$

$$a_{-1} = \frac{3}{2} - \frac{1}{2}j = \frac{3}{2} + \frac{1}{2}j$$

$$a_0 = 1 \quad \leftarrow \text{don't forget}$$

$$a_1 = \frac{3}{2} + \frac{1}{2}j = \frac{3}{2} - \frac{1}{2}j$$

$$a_2 = \frac{1}{2}j$$

I see now

(27)

b)  $x[n] = 5 \cos(6\pi n + \pi) + 7 \cos(3\pi n)$

find smallest  $N$  where periodic  
rewrite for that

$a_{\pm 6} = 5 \cdot \frac{1}{2}$

$a_{\pm 3} = 7 \cdot \frac{1}{2}$

$N = 2$  here

$a_0 = -5$

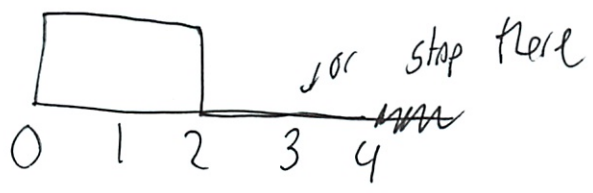
$a_1 = \frac{7}{2}$

$\cos(\pi)$  is just 1

period =  $\frac{2\pi}{2} = \frac{2\pi}{6\pi} = \frac{1}{3}$  iff

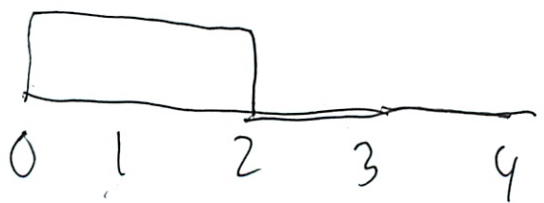
c)  $x[n]$  is square wave w/ period  $N=4$

$x[0] = 1$     $x[1] = 1$     $x[2] = 0$     $x[3] = 0$



$a_k = \frac{1}{N} \sum_{n \in \mathbb{Z}} x[n] e^{-jk \frac{2\pi}{N} n}$

do it manually if need to



$a_{-2} = 0 = \frac{1+j}{4}$

$a_1 = \frac{1+j}{4}$

no think go to  $a_0 = 1$

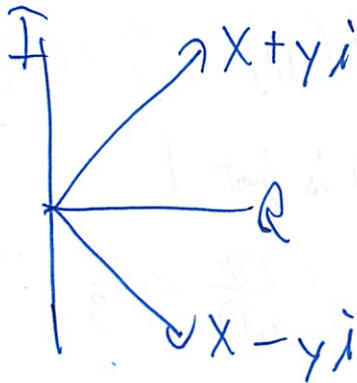
but I don't know how to get  $a_n$     $a_1 = \frac{1-j}{4}$

remember  $a_k$  is complex conj of  $a_{-k}$  when  $x[n]$  is real

Complex conj = pair of complex #

same real part

flip signs on imag part



(28)

$$d) x[n] = \cos\left(\frac{2\pi n}{3}\right) \circ \sin\left(\frac{2\pi n}{9}\right)$$

Just look at ans

Smallest  $N$  where period is is 9

↑ Where/how?

$$\frac{2\pi}{\frac{2}{3}} = \frac{2\pi \cdot 3}{2} = \frac{6\pi}{2} = 3\pi$$

$$a_{-4} = \frac{j}{4}$$

$$a_k = 0$$

$$\frac{2\pi}{\frac{2}{9}} = \frac{18\pi}{2} = 9\pi$$

$$a_{-2} = -\frac{j}{4}$$

$$\text{hank } \frac{2\pi}{N} \cdot k$$

if think I will skip this

$$a_0 = 0$$

$$a_2 = \frac{j}{4}$$

so  ~~$N=2$~~  / but not what got before

$$a_4 = -\frac{j}{4}$$

I'm just not seeming to learn this part

e)  $x[n]$  has 1 non-zero value per period  
for  $x[m] \neq 0$  for  $m$  and 0 otherwise

Mag of  $a_k$

(which  $a_k$ 's)

$$|a_k| = \left| \frac{1}{N} \sum_{[n]} x[n] e^{-jk \left(\frac{2\pi}{N} n\right)} \right|$$

$$= \frac{1}{N} |x[m]| |e^{-jk \frac{2\pi}{N} n}|$$

$$= \frac{|x[m]|}{N}$$

Since  $|e^{j\ell}|$  is 1 for any  $\ell$

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4. If  $x[n]$  is real, even ~~and~~ periodic w/ period  $N$   
Show that all  $a_k$  are real

$$a_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk \frac{2\pi}{N} n}$$

Noting that  $x[1] = x[-1], x[2] = x[-2], \dots$   
if  $N$  is odd, can rewrite

$$a_k = \frac{1}{N} \left[ x[0] + x[1] \left( e^{jk \frac{2\pi}{N}} + e^{-jk \frac{2\pi}{N}} \right) + x[2] \left( e^{jk \frac{2\pi}{N} 2} + e^{-jk \left( \frac{2\pi}{N} \right) 2} \right) + \dots \right]$$
$$= \frac{1}{N} \left[ x[0] + 2 \cos \left( \frac{2\pi k}{N} \right) x[1] + 2 \cos \left( \frac{4\pi k}{N} \right) x[2] + \dots \right]$$

which is a sum of real #s

If  $N$  is even there will be ~~an~~ - index in eq  $n = -\frac{N}{2}$  that is not matched w/  $\oplus$  counterpart

$$\text{But } x[-N/2] e^{-jk \left( \frac{2\pi}{N} \right) \left( -\frac{N}{2} \right)}$$
$$= x[-N/2] e^{jk \pi}$$
$$= x[-N/2] \cdot (-1)^k$$

which is real

did not follow!

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5. Given  $a_k$  for a seq  $x[n]$  compute  $b_k$  for  $w[n] = x[n-d]$

Just shift?

$$b_k = \frac{1}{N} \sum_{n=\langle N \rangle} w[n] e^{-jk \left(\frac{2\pi}{N}\right) n}$$

$$= \frac{1}{N} \sum_{n=\langle N \rangle} x[n-d] e^{-jk \left(\frac{2\pi}{N}\right) n}$$

This time (n indices of summation from  $d$  to  $d+(N-1)$ )

$$b_k = \frac{1}{N} \sum_{d \leq n \leq d+(N-1)} x[n-d] e^{-jk \left(\frac{2\pi}{N}\right) n}$$

$$= \text{" " } x[n] e^{-jk \left(\frac{2\pi}{N}\right) n+d}$$

$$= \text{" " } x[n] e^{-jk \left(\frac{2\pi}{N}\right) n} + e^{-jk \frac{2\pi}{N} d}$$

$$= e^{-jk \frac{2\pi}{N} d} \cdot \frac{1}{N} \sum_{0 \leq n \leq (N-1)} x[n] e^{-jk \frac{2\pi}{N} n}$$

$$= a_k e^{-jk \left(\frac{2\pi}{N}\right) d}$$

(31) I know how to do certain things + useless at other stuff

6. Consider an LTI system w/  $h[n]$

a) Give  $H(e^{j2})$  in terms of  $h[n]$

$$= \sum_m h[m] e^{-j2m} \quad \checkmark$$

b) If  $h[0]=1$   $h[1]=0$   $h[2]=1$   $h[n]=0$  for other  $n$

What is  $H(e^{j2})$

$$= e^{j2 \cdot 0} + 0 e^{j2 \cdot 1} + 1 \cdot e^{j2 \cdot 2}$$

$$= 1 + 0 + e^{j2 \cdot 2} \quad \checkmark$$

$$= 1 + \cos(2\ell) + \sin(-2\ell)j$$

*(neg since back in time)*

I don't know how much this is helping - till look at ans



(32)

c) Let  $h[n]$  be b

$$x[n] = \cos \omega n$$

Is there a  $\omega$  so  $y[n] = 0$  for all  $n$   $0 \leq \omega \leq \pi$

So

~~$$0 = 1 + \cos(2\omega)$$~~

$$0 = 1 + 2x$$

$$-1 = 2x$$

$$x = -\frac{1}{2}$$

$$-\frac{1}{2} = \cos(\omega) \quad \uparrow \text{imag is } 0$$

$$\frac{2\pi}{3}$$

$\uparrow$  only top half of circle

$\frac{\pi}{2}$  should be

d) Find max mag of  $y[n]$  if  $x[n] = \cos(\pi \frac{n}{4})$

What does this mean again?

$$\omega = \frac{\pi}{4}$$

$$= 1 + \cos\left(2 \frac{\pi}{4}\right) + \sin\left(2 \frac{\pi}{4}\right) i$$

$$= 1 + 0 + i$$

$$= 1 + i$$

- what does max mag mean?

$$= \sqrt{2}$$

(33)

e)  $h[n]$  as before. Find max mag of  $y[n]$  if  $x[n] = \cos\left(\frac{-\pi}{2}n\right)$

So same, but  $\ell = -\frac{\pi}{2}$

$$= 1 + \cos\left(2 \cdot -\frac{\pi}{2}\right) + \sin\left(2 \cdot -\frac{\pi}{2}\right)j$$

$$= 1 + \cos(-\pi) + \sin(-\pi)j$$

$$= 1 + \cos(\pi) + \sin(\pi)j$$

$$= 1 + -1 + 0$$

$$= 0 \quad \checkmark$$

(34)

7. 2 filters

4 systems

a) what is  $h[n] = \alpha \delta[n] - h_1[n]$

and what is  $\alpha$

- no idea

$\delta$  is like 1?

C?

D - I'm just pattern matching

B has only freq response w/ same values at  $\omega, \pm\pi, \pm 2\pi$

$\alpha = 2$  as  $H_1(e^{j0}) = 2$  But  $H_0(e^{j0}) = 0$

B)  $h[n] = \sum_m h_1[m] h_2[n-m]$  for  $m = 0$  to  $n$

MA since  $H_1(e^{j\omega}) H_2(e^{j\omega}) = H(e^{j\omega})$

so  $|H(e^{j\omega})| = 0$  when  $H_1(e^{j\omega}) = 0$  or  $H_2(e^{j\omega}) = 0$

(35)

8. Last problem!

3 filters as shown

a) What freq response is this?

What is max  $\Omega$